

Intelligent Antenna Sharing in Cooperative Diversity Wireless Networks

Ph.D. Thesis Defense
June 2005

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Viral Communications Group, MIT Media Laboratory

Thesis Committee

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Associate Professor, MIT Media Lab.
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Motivation and Inspirations

You are (probably) here because you have all experienced:

- *bad reception...*
- *battery problems...*
- *no connectivity during large gatherings (4th of July problem!)*...

Could we fix all the above problems?

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- Gupta and Kumar IT 2000 result: local communication helps...
- Multiple Antennas at each radio help...

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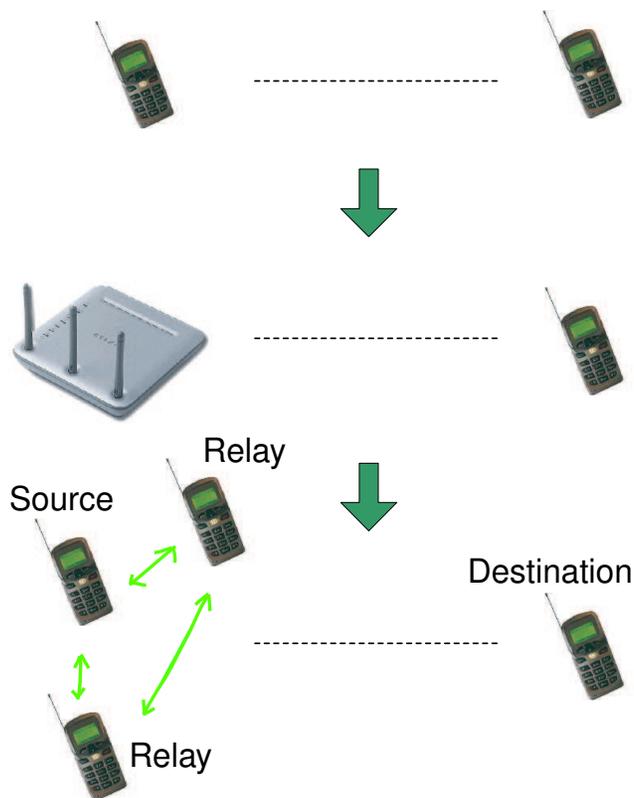
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Could we fix all the above problems?

Inspirations:

- Gupta and Kumar IT 2000 result: local communication helps...
- Multiple Antennas at each radio help...
- Could we merge the two above? More users $\stackrel{?}{=} better$ wireless communication?

Additional Problem Constraint: Low Complexity and Implementation

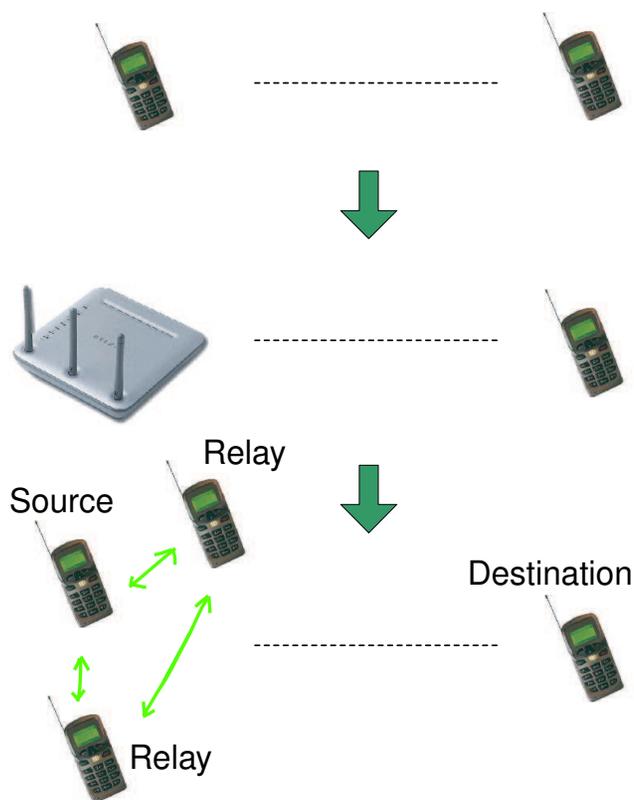


In general, multi-antenna systems increase:

- reliability (diversity gain).
- spectral efficiency bps/Hz (multiplexing gain)

- Explore multiple antennas in the *Relay* channel, via *cooperative* relays.

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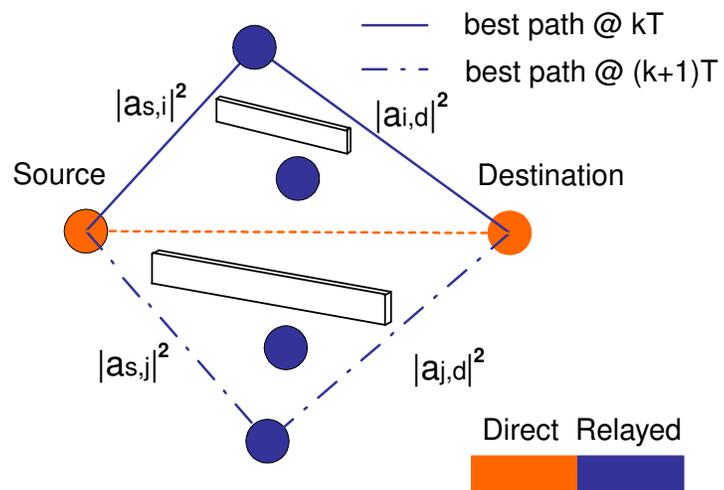


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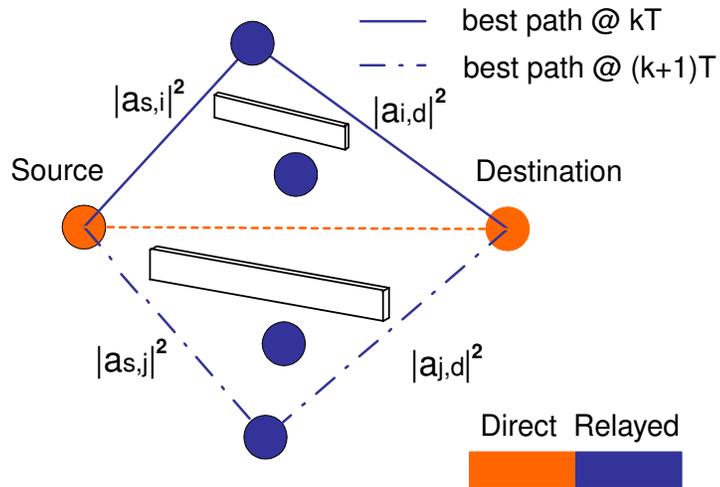
- Explore multiple antennas in the *Relay* channel, via *cooperative* relays.
- IMPLEMENTATION TODAY, with existing RF-front ends.

Main Difficulties



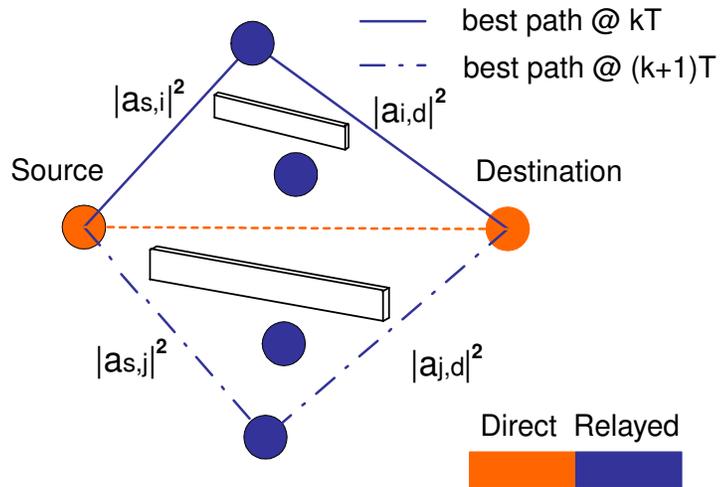
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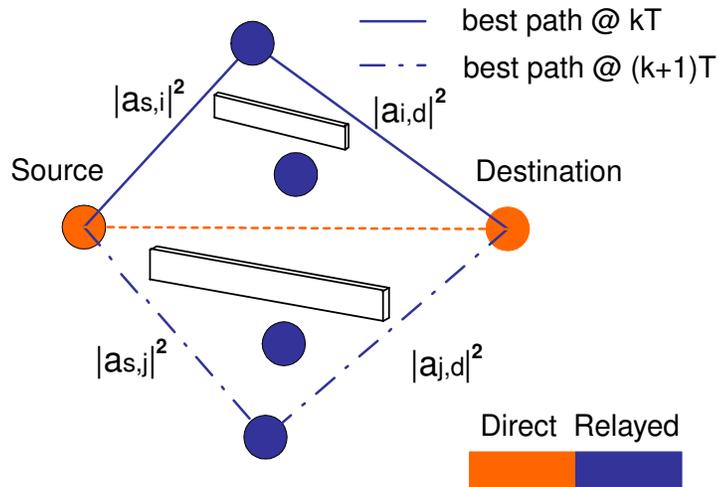
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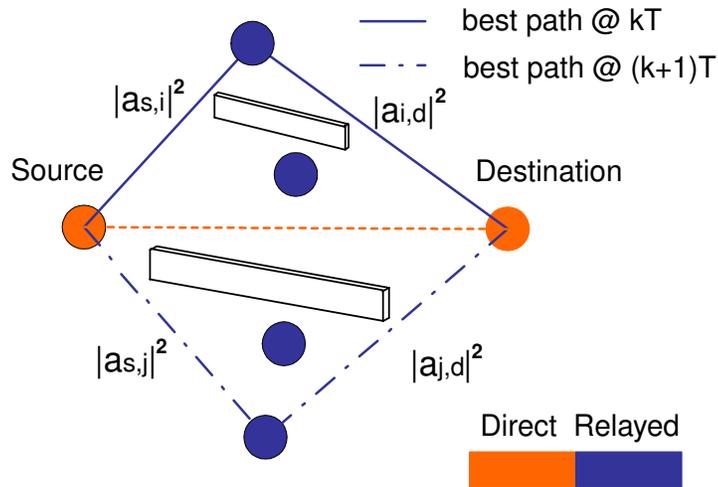
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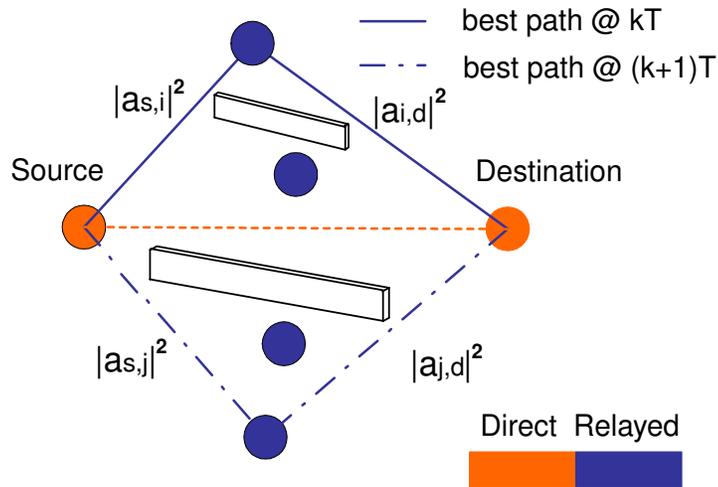
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- Radio transceiver complexity.

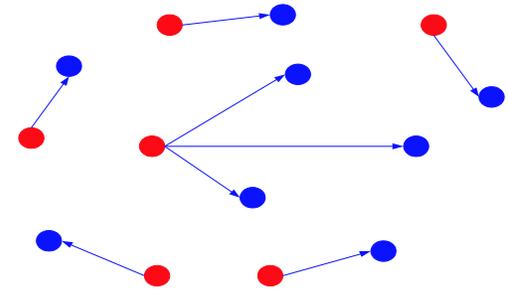
Outline

- Assumptions and Background
- Approach
- Performance
- Implementation Example
- Relevant Technologies
- Conclusion
- Acknowledgements

Assumptions and System Model

Inline with prior art in the field:

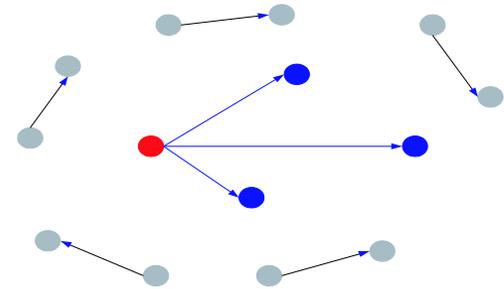
- Half-duplex radios.
- Simple RF-front ends:
 - Half-duplex radios.
 - No rate adaptation (no CSI at the source).
 - No phased arrays (No *beamforming*).
- $y_d = a_{sd} x_s + n_d$.
- Neighboring *interfering* streams: noise.
- (Mostly) Rayleigh fading. $\mathcal{E}[|a_{sd}|^2] = 1/d^v$
- Slow Fading (most difficult communication problem).



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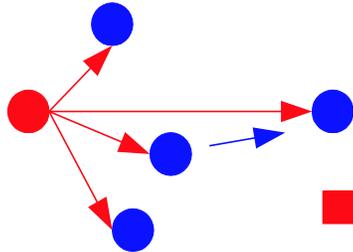
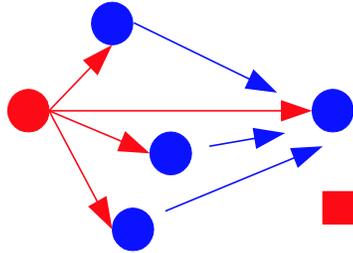
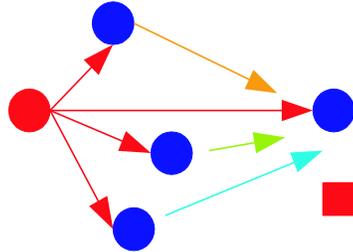
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Approaches



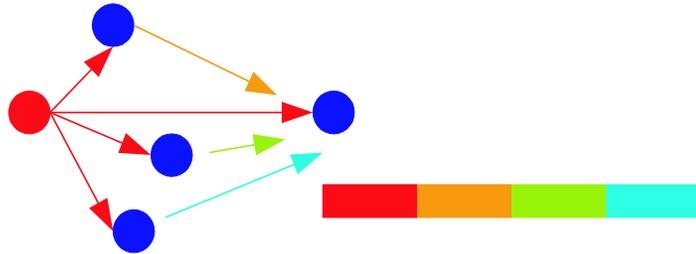
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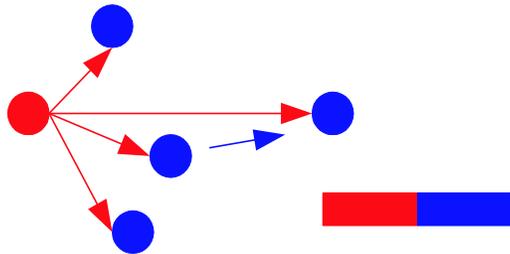
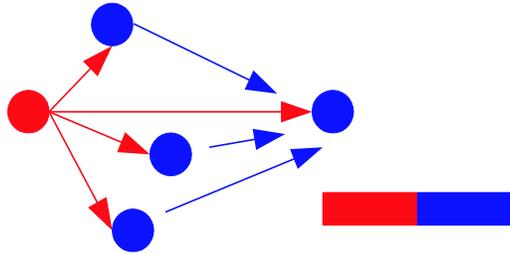
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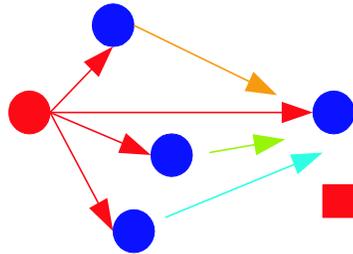
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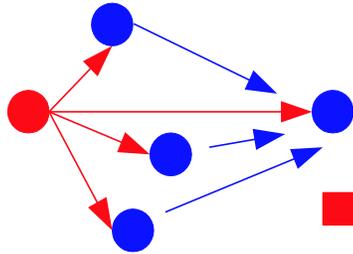
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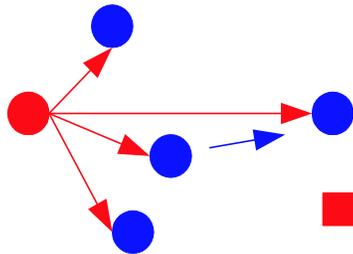
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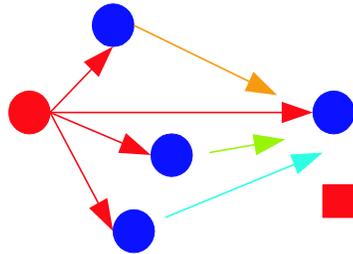
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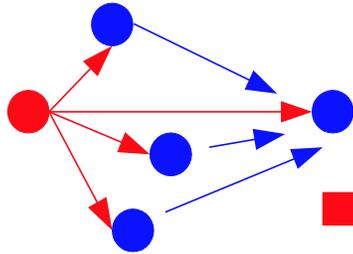
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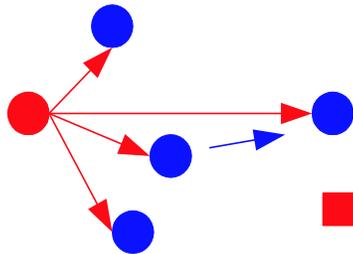
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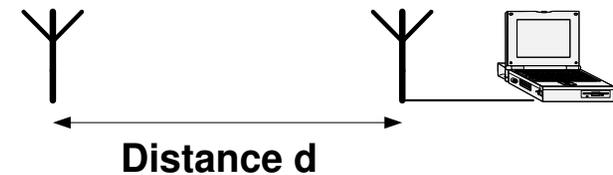
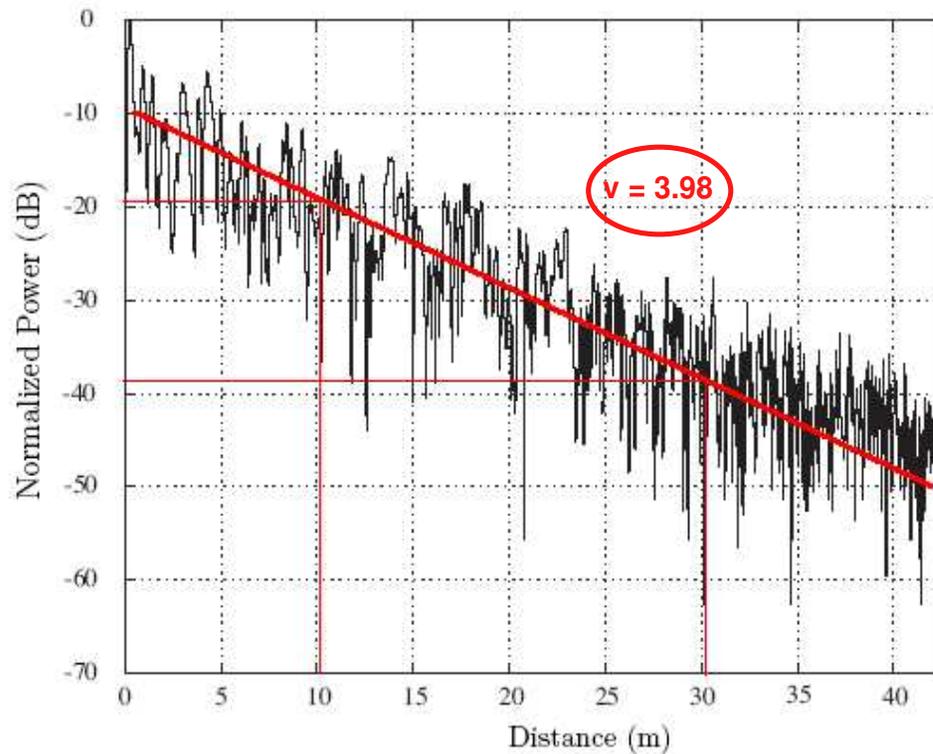
● Our Approach.

- *Proactive* single relay selection.
- *Instantaneous* channel conditions (instead of average).

Outline

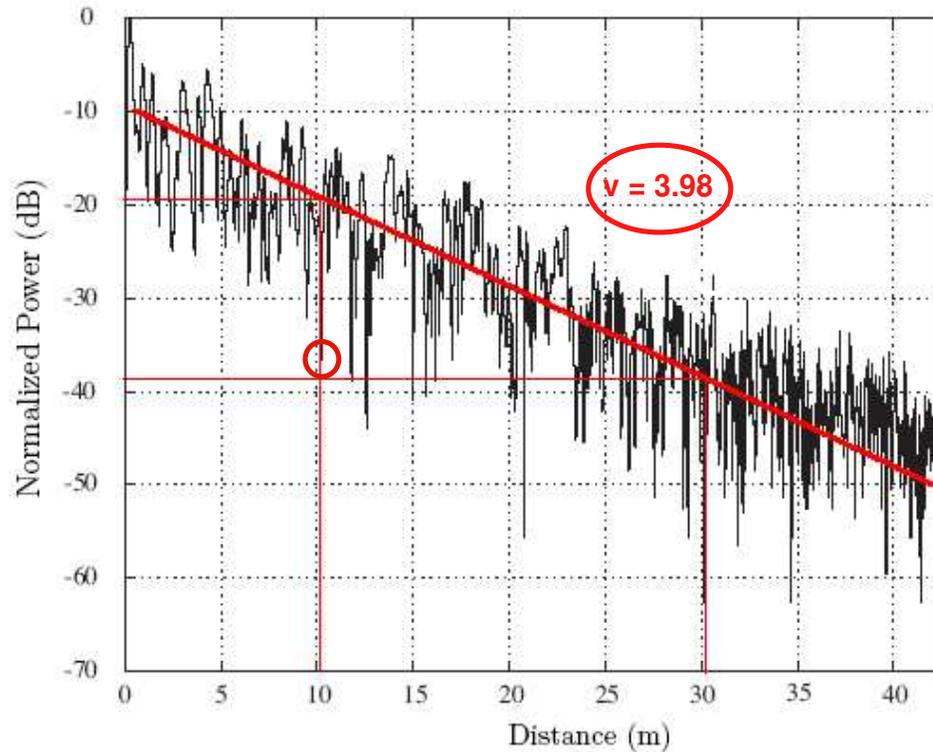
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Wireless Channel Observations



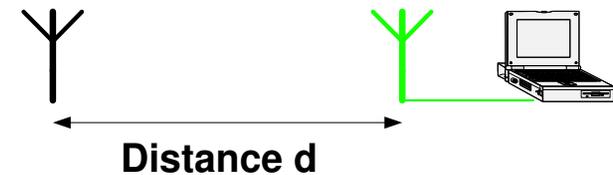
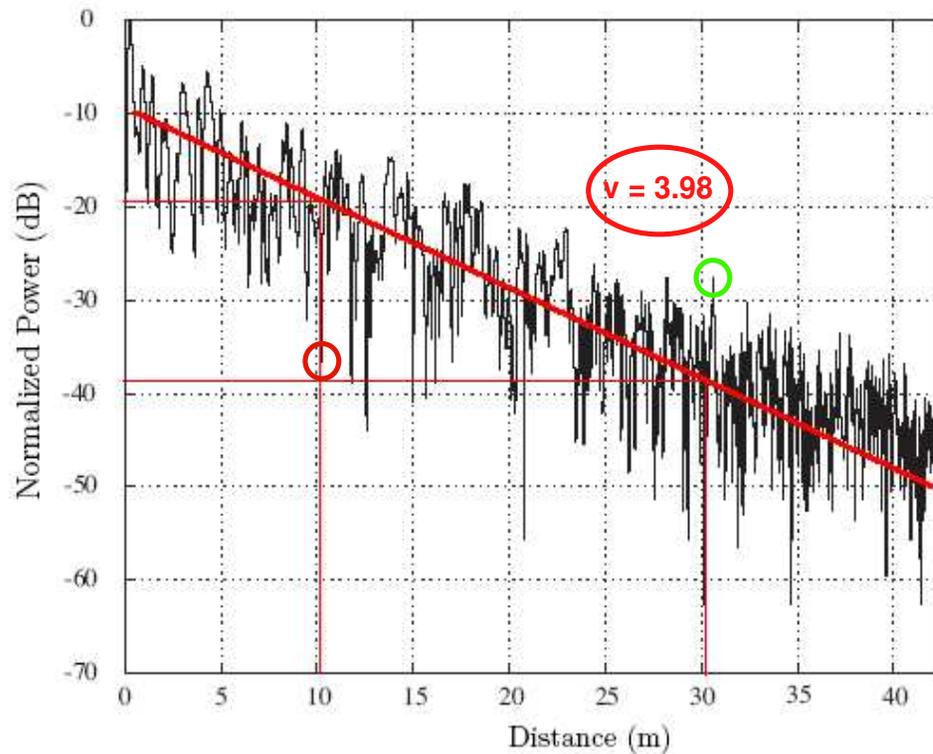
- Receiver cares about signal strength (not distance).
- Selection based on distance or *average* SNR... is suboptimal.
- *Instantaneous* channel conditions matter!

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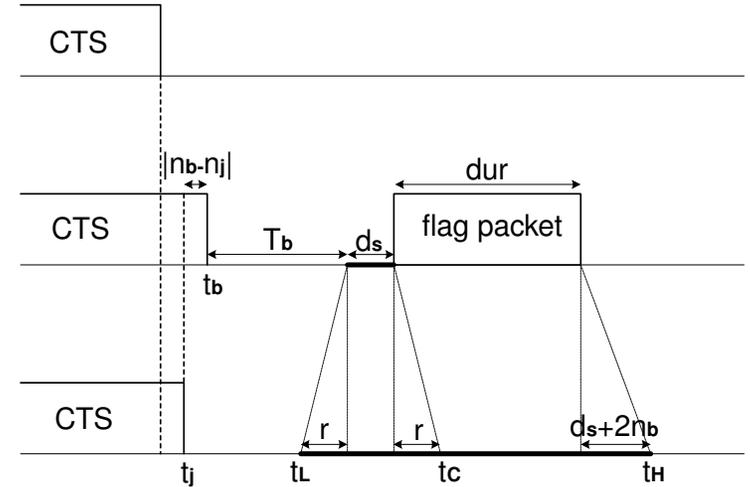
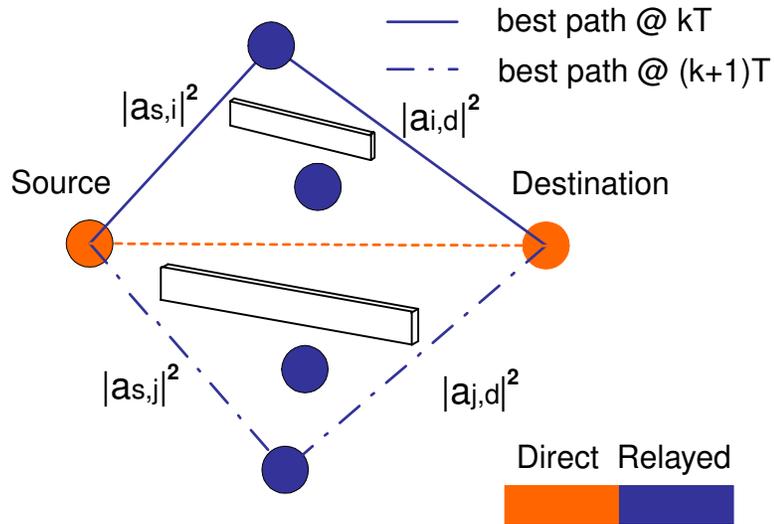
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Our Approach: *Opportunistic Relaying*



$$Policy I : h_i = \min\{|a_{si}|^2, |a_{id}|^2\} \quad Policy II : h_i = \frac{2}{\frac{1}{|a_{si}|^2} + \frac{1}{|a_{id}|^2}} = \frac{2 |a_{si}|^2 |a_{id}|^2}{|a_{si}|^2 + |a_{id}|^2}$$

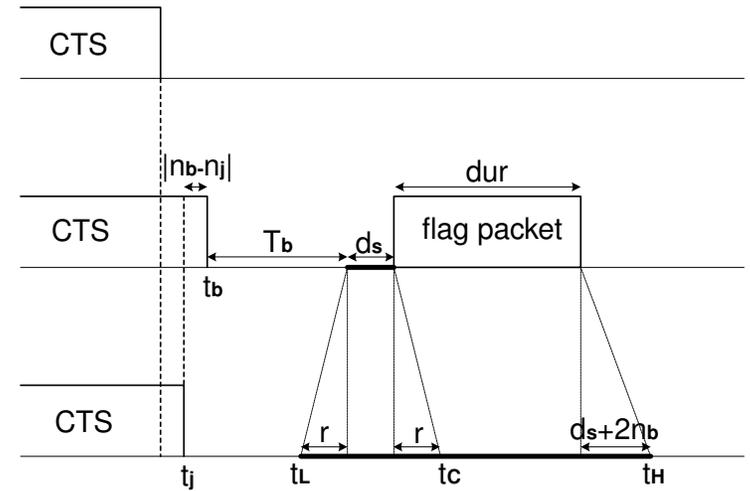
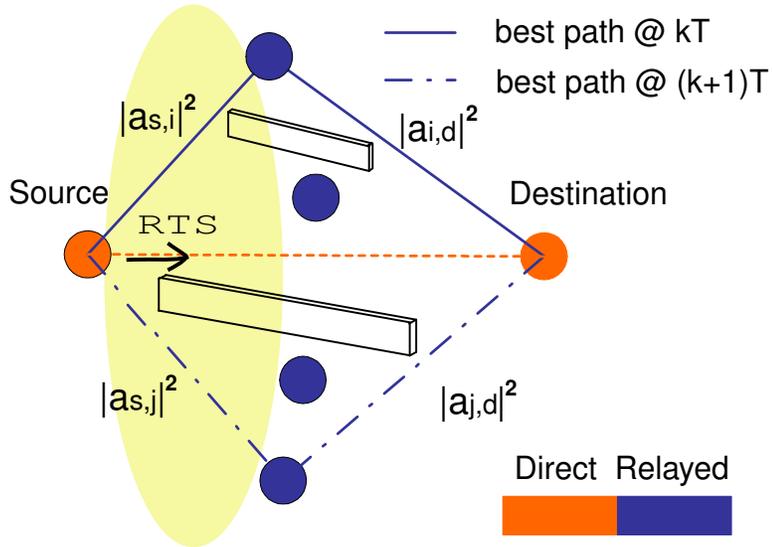
$$T_i = \frac{\lambda}{h_i} \quad (1)$$

Here λ has the units of time. For the discussion in this work, λ has simply values of $\mu secs$.

$$h_b = \max\{h_i\}, \iff \quad (2)$$

$$T_b = \min\{T_i\}, i \in [1..M]. \quad (3)$$

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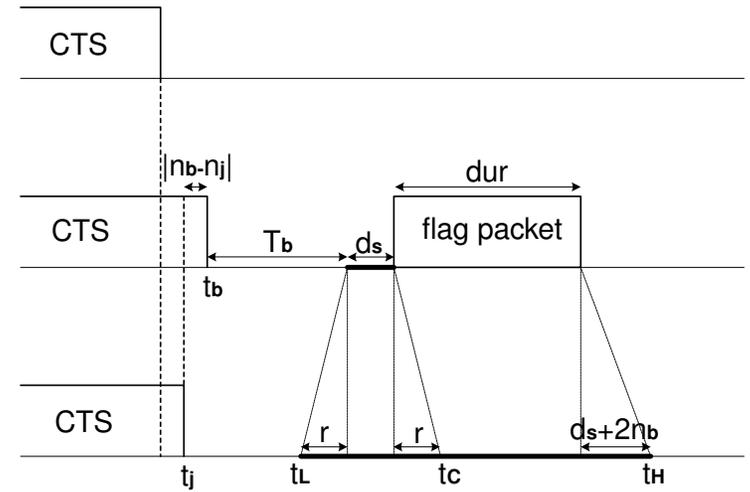
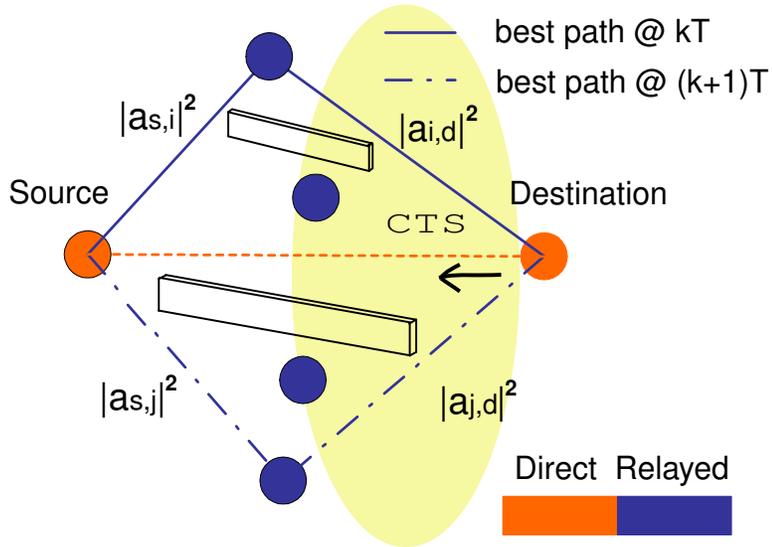
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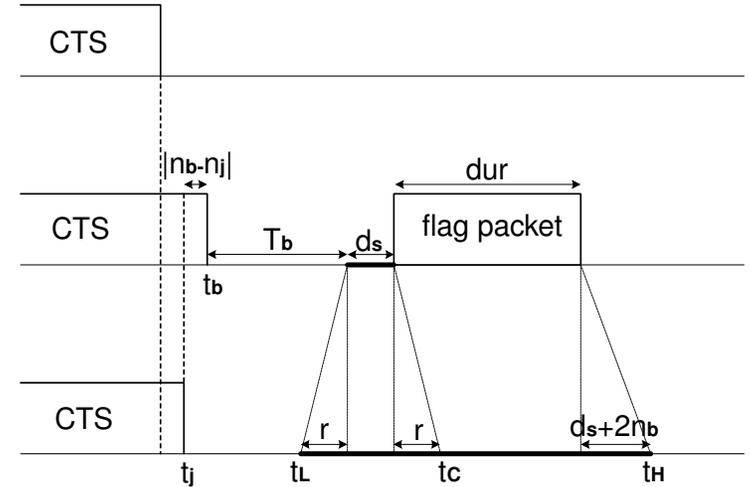
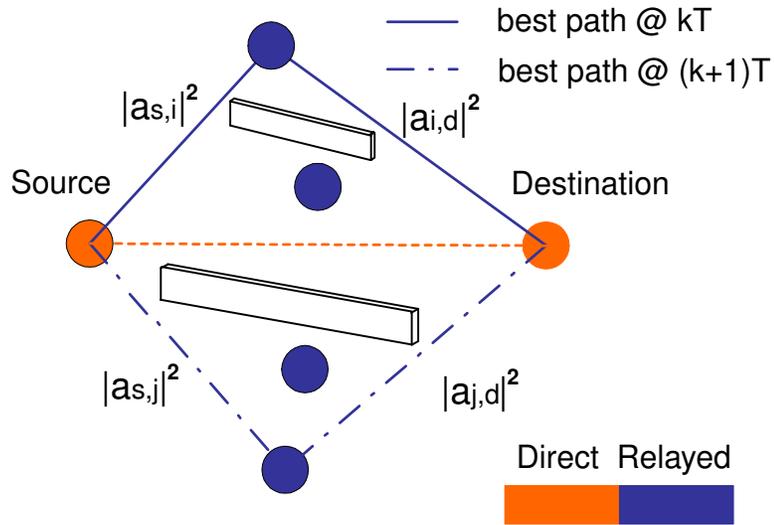
$$T_i = \frac{\lambda}{h_i} \quad (7)$$

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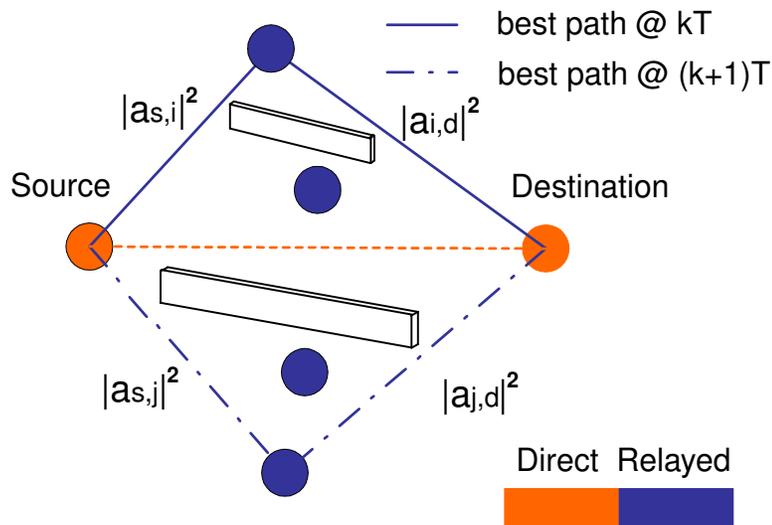
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Discussion: a note on CSI and time synchronization

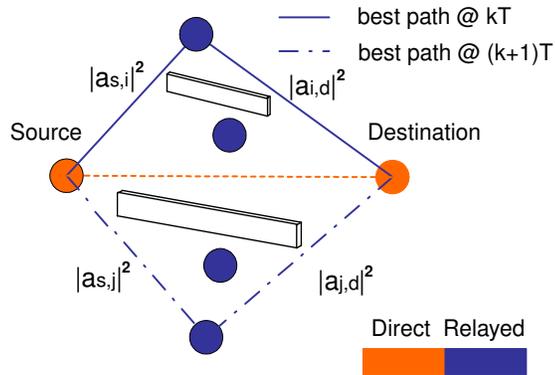


- RTS/CTS exchange is only needed at the relays to estimate uplink/downlink channel.
- CTS reception is not exploited at the source.
- No *beamforming* or rate adaptation at the relays.
- No need for an explicit time sync protocol.
- It is a multi-hop scheme.
- We do know that the term "Opportunistic" has been used before...

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Outage Performance (1)



- Outage event between source s and destination d :

$$\log(1 + |a_{sd}|^2 SNR) \leq \rho \Leftrightarrow |a_{sd}|^2 \leq (2^\rho - 1)/SNR \Leftrightarrow \gamma_{sd} \leq \Theta$$

- "Best" opportunistic relay is chosen, according to *instantaneous*, end-to-end channel conditions:

$$b = \underbrace{\arg \max}_i \{ \min\{\gamma_{si}, \gamma_{id}\} \}, \quad i \in [1..M] \quad (13)$$

- Probability of outage via "best" relay:

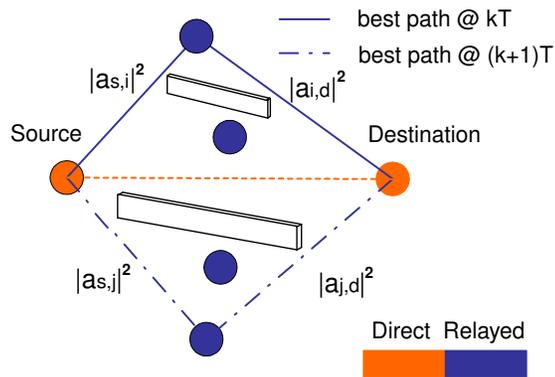
$$P_r(\gamma_{sb} < \Theta_2 \cup \gamma_{bd} < \Theta_2), \quad \Theta_2 = 2(2^{2\rho} - 1)/SNR \quad (14)$$

Outage Performance (2)

- The above outage probability of opportunistic relaying is calculated for the case of Rayleigh Fading:

$$Pr(\gamma_{sb} < \Theta_2 \cup \gamma_{bd} < \Theta_2) = \prod_{i=1}^M (1 - \exp(-\Theta_2 (\frac{1}{\bar{\gamma}_{si}} + \frac{1}{\bar{\gamma}_{id}}))) \quad (15)$$

- Taking into account the direct path between source and destination, the overall outage probability becomes:



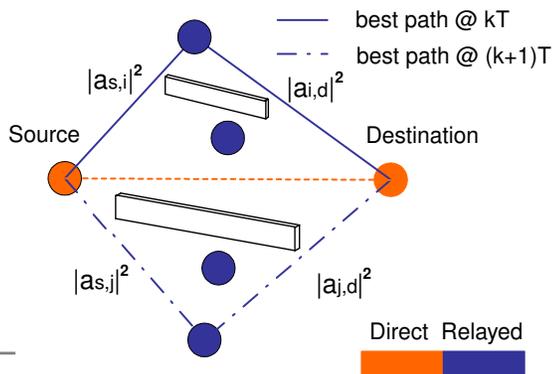
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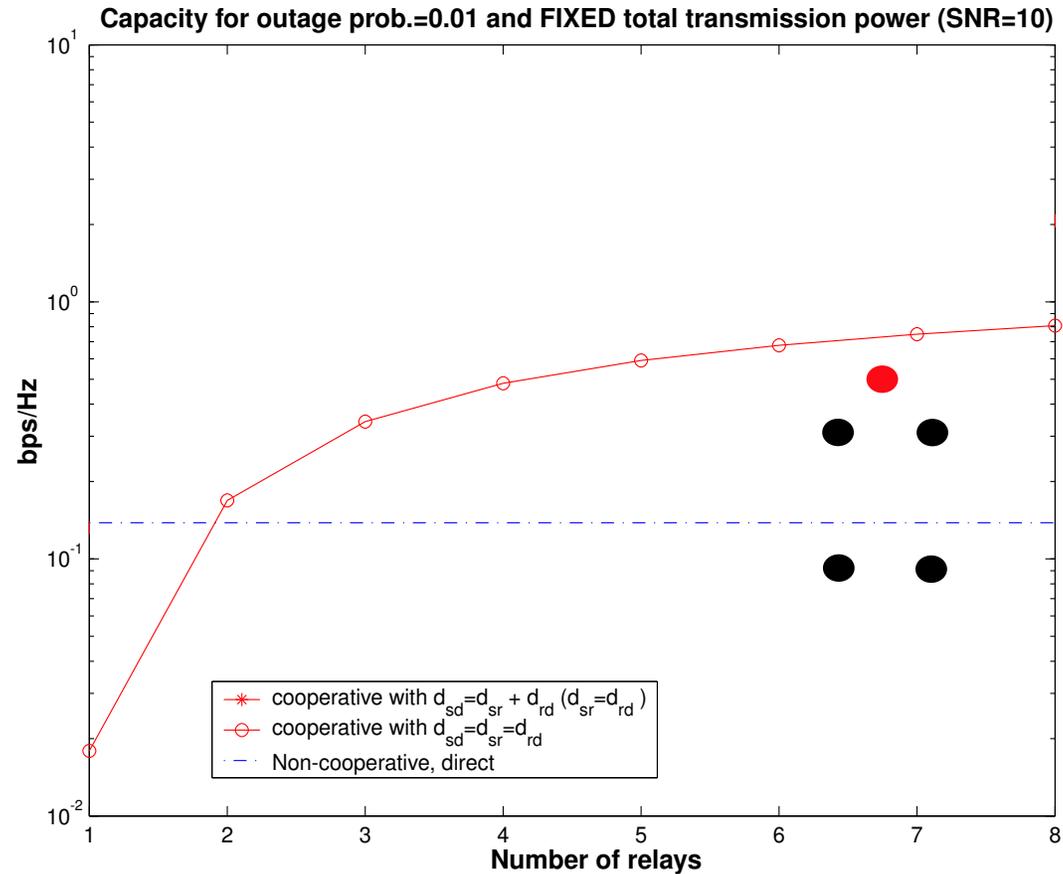
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- Taking into account the direct path between source and destination, the overall outage probability becomes:

$$P_r^{out} = \underbrace{(1 - \exp(-\Theta_2/\bar{\gamma}_{sd}))}_{direct} \underbrace{\prod_{i=1}^M (1 - \exp(-\Theta_2 (\frac{1}{\bar{\gamma}_{si}} + \frac{1}{\bar{\gamma}_{id}})))}_{relaying} \quad (17)$$

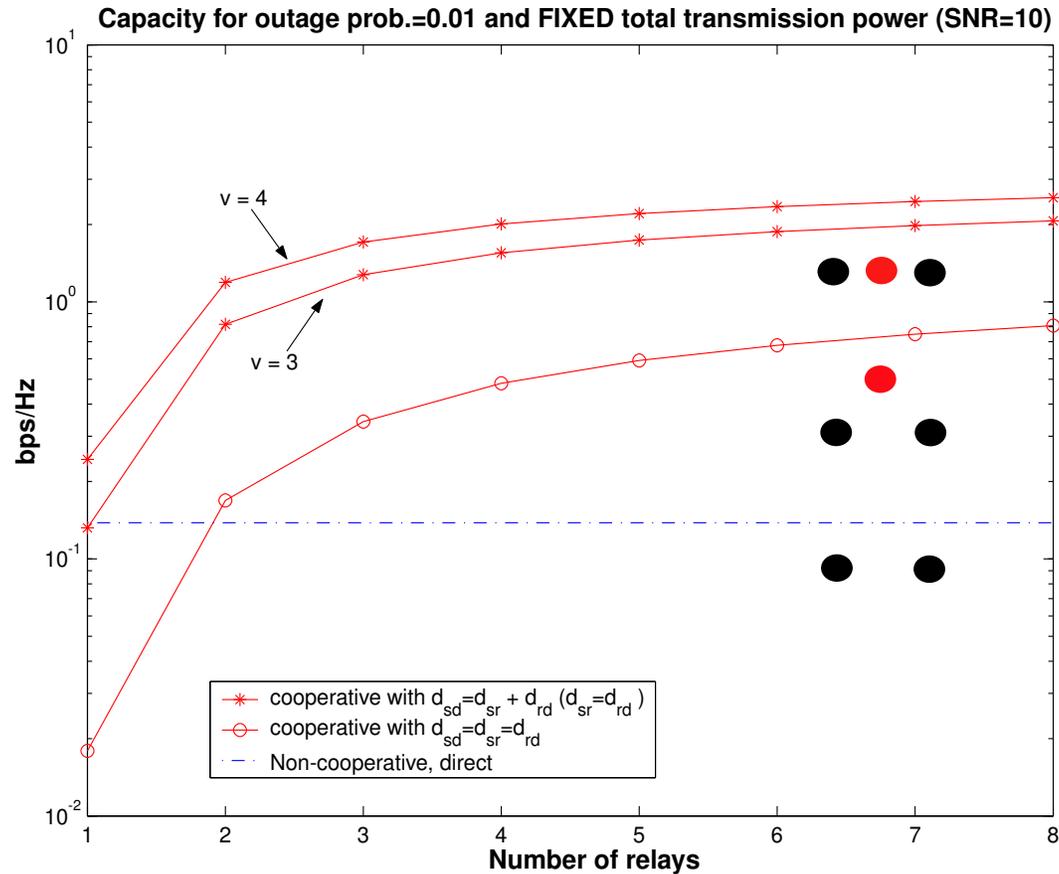


Outage Performance (3)



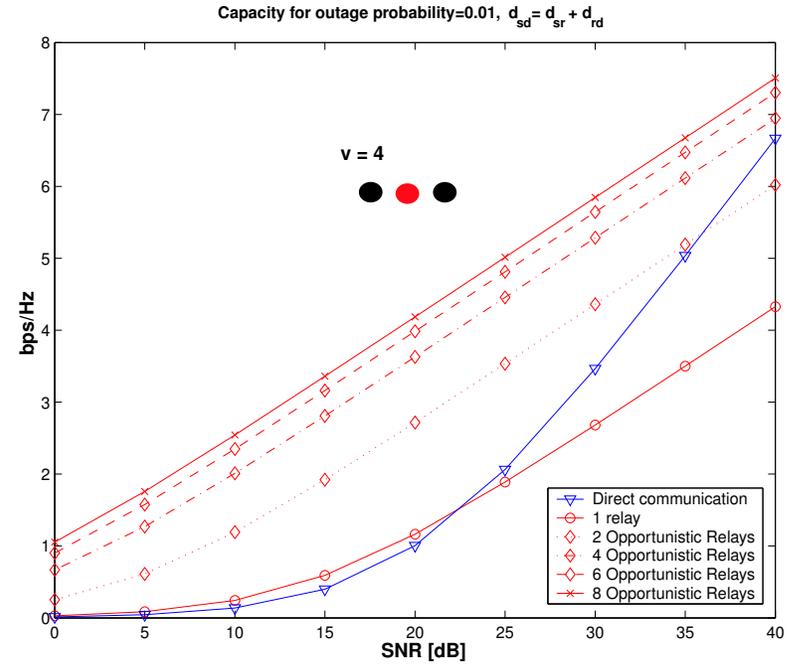
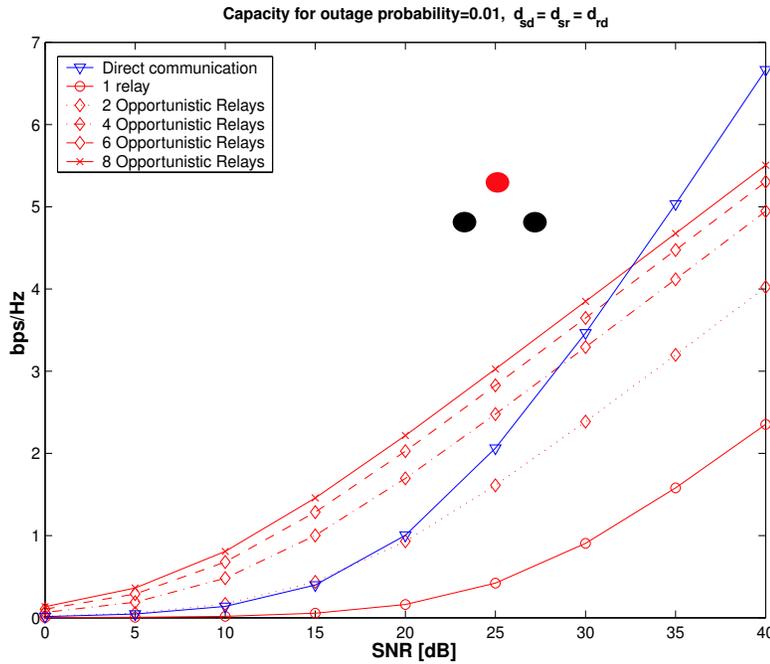
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- Opportunistic relays do help, even under a total tx power constraint!

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Outage Performance (4)



 $P_{r^{out}} = \delta.$

$$\rho_{opport} = \frac{1}{2} \log_2(1 - \ln(1 - \delta^{1/M}) \frac{SNR}{2} \bar{\gamma}_{sid}) \quad (18)$$

$$\rho_{direct} = \log_2(1 - \ln(1 - \delta) SNR \bar{\gamma}_{sid}) \quad (19)$$

Diversity-Multiplexing Tradeoff (1)

$$d \triangleq - \lim_{SNR \rightarrow \infty} \frac{\log P_e(\rho)}{\log SNR} \qquad r \triangleq \lim_{SNR \rightarrow \infty} \frac{\rho(SNR)}{\log SNR}$$

- Diversity-Multiplexing Gain tradeoff tool averages out geometry.
- cooperative diversity \neq multihop communication. This tool can reveal associated gains/losses.
- **Theorem 0:** The achievable diversity multiplexing tradeoff for the decode and forward strategy with M intermediate relay nodes is given by $d(r) = (M + 1)(1 - 2r)$ for $r \in (0, 0.5)$.

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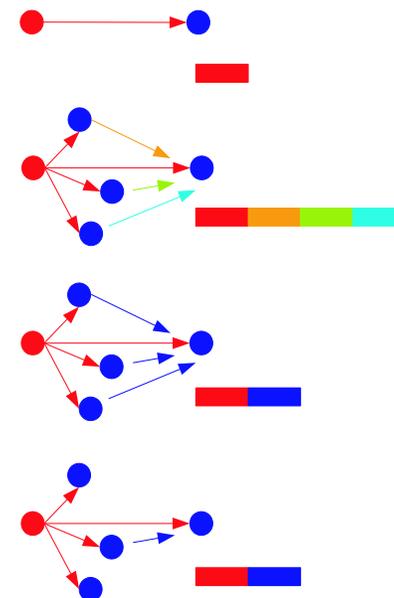
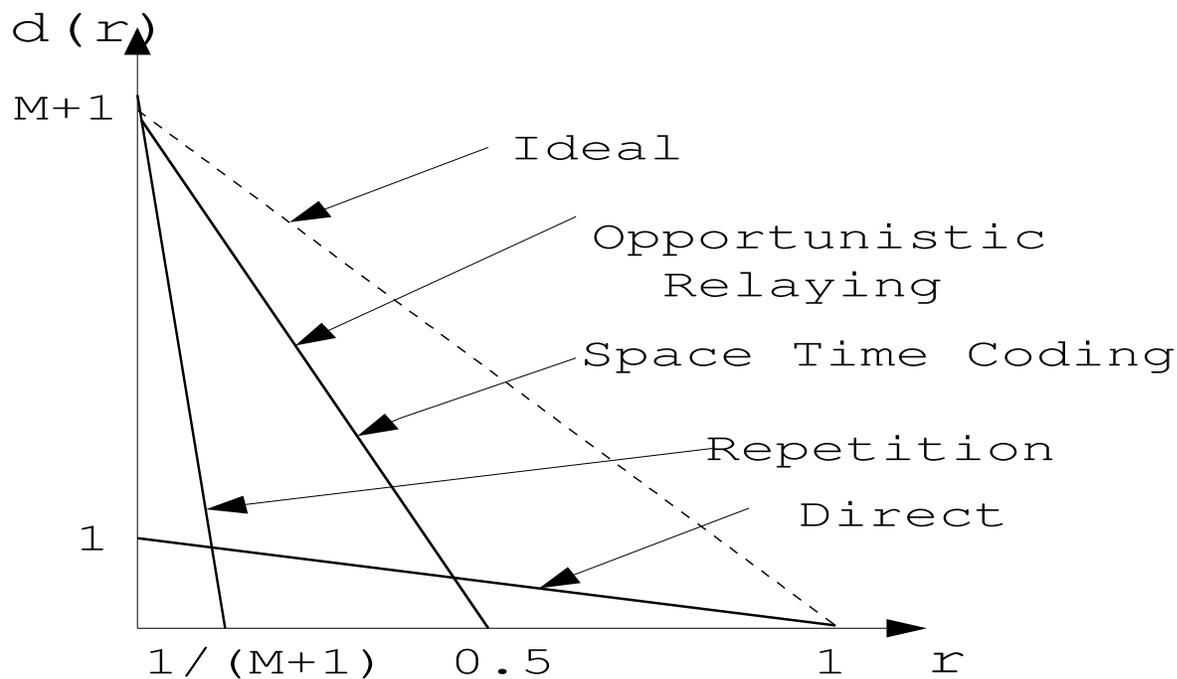
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- **Theorem 1*:** Under opportunistic relaying, the decode and forward protocol with M intermediate relays achieves the same diversity multiplexing tradeoff, as in Theorem 0.
- **Theorem 2*:** Opportunistic amplify and forward achieves the same diversity multiplexing tradeoff stated in Theorem 0.

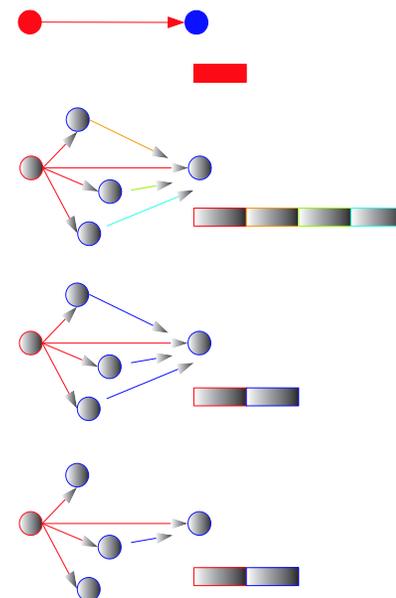
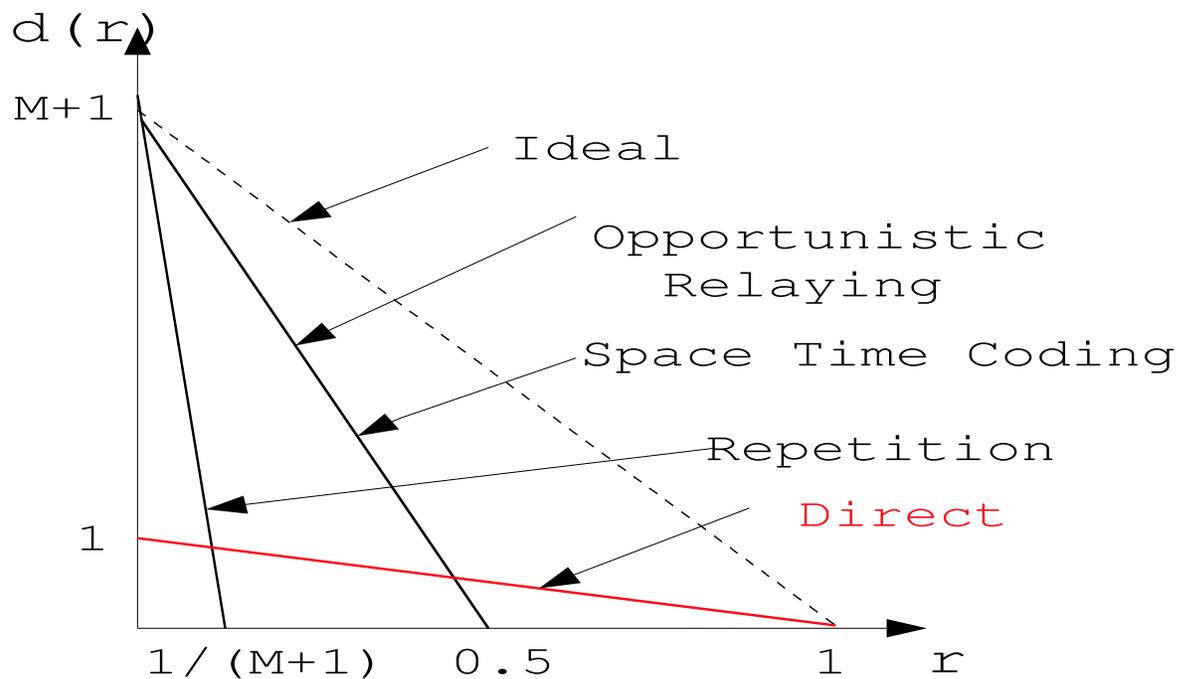
*: In cooperation with Ashish Khisti.

Diversity-Multiplexing Tradeoff (2)



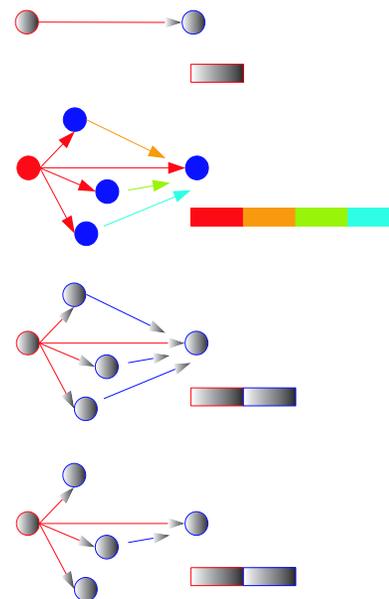
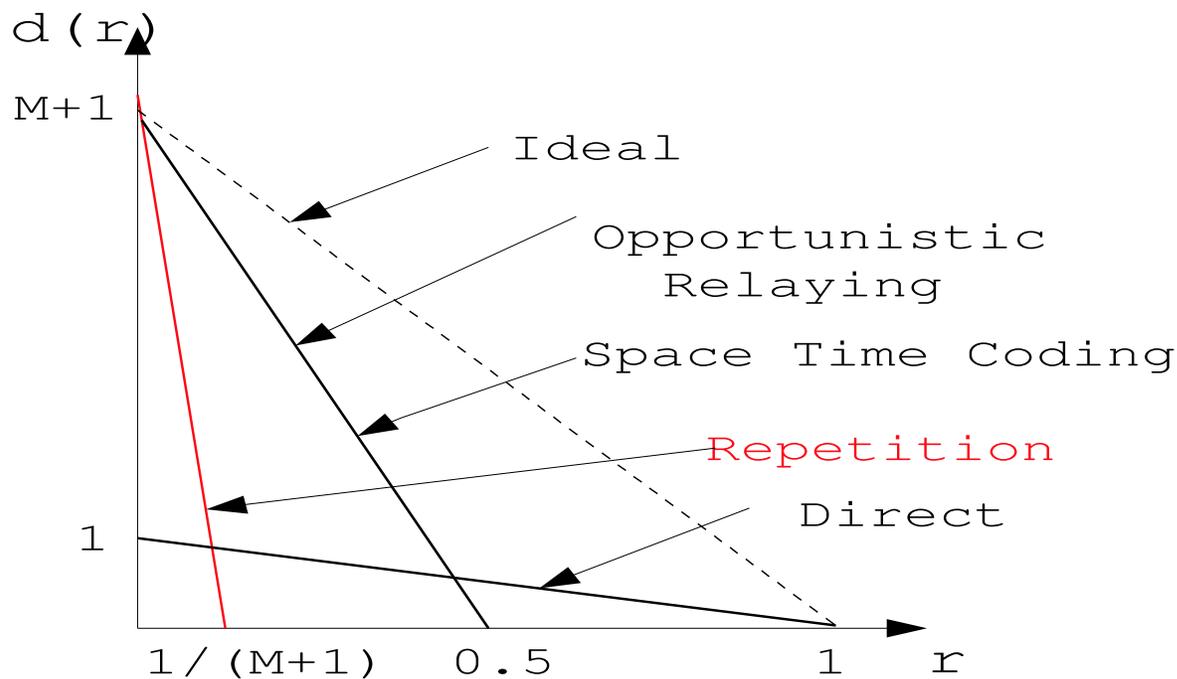
- *Opportunistic*, single relay selection is as good as space-time coding simultaneous transmissions!
- This result holds for decode/forward as well as amplify/forward!

Diversity-Multiplexing Tradeoff (2)



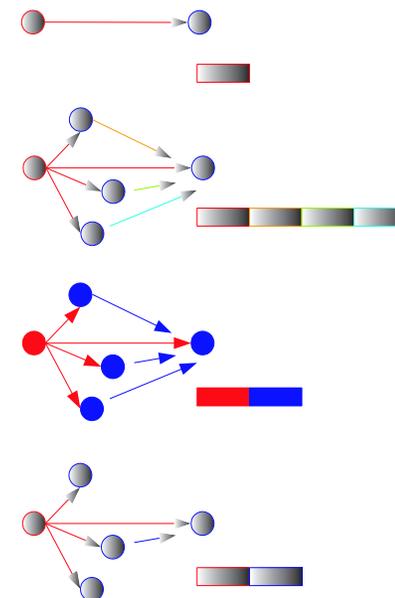
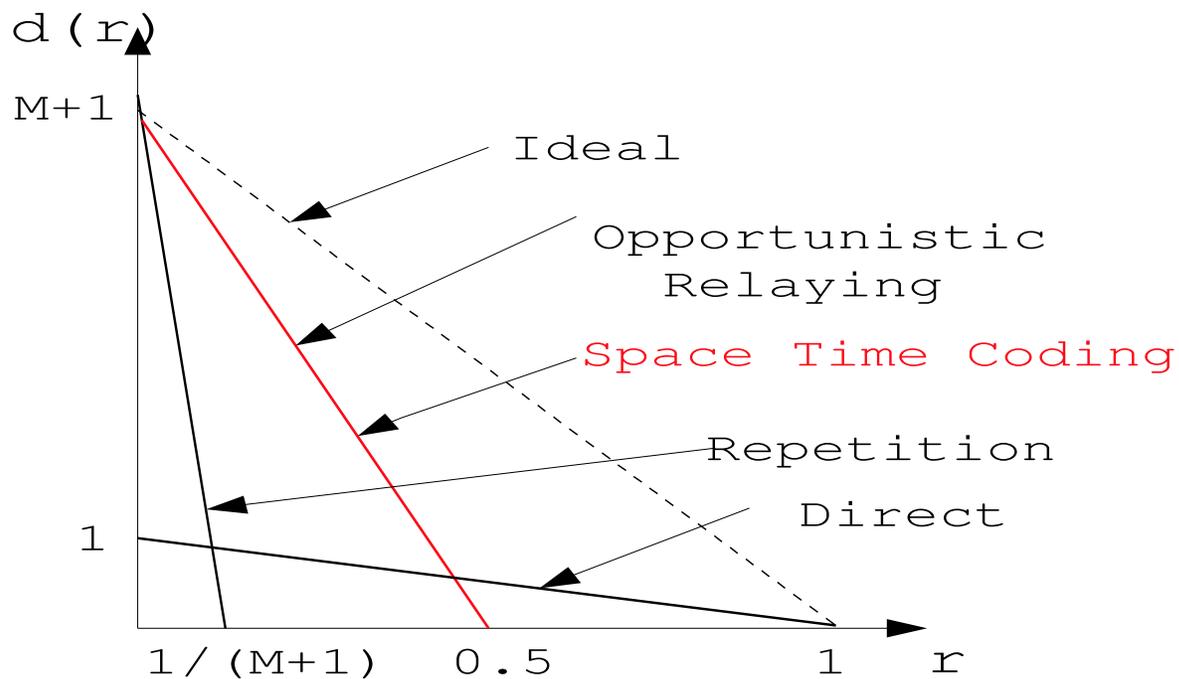
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- This result holds for decode/forward as well as amplify/forward!

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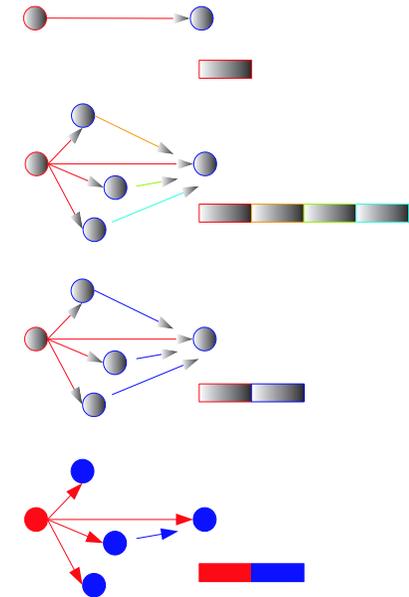
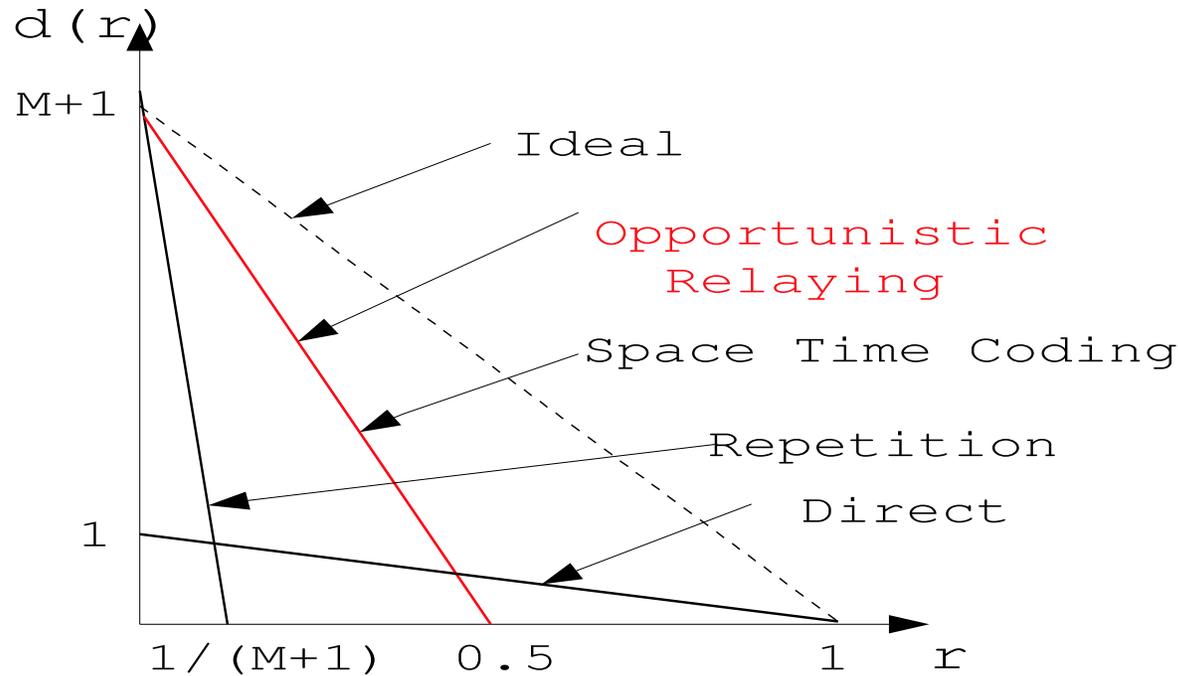
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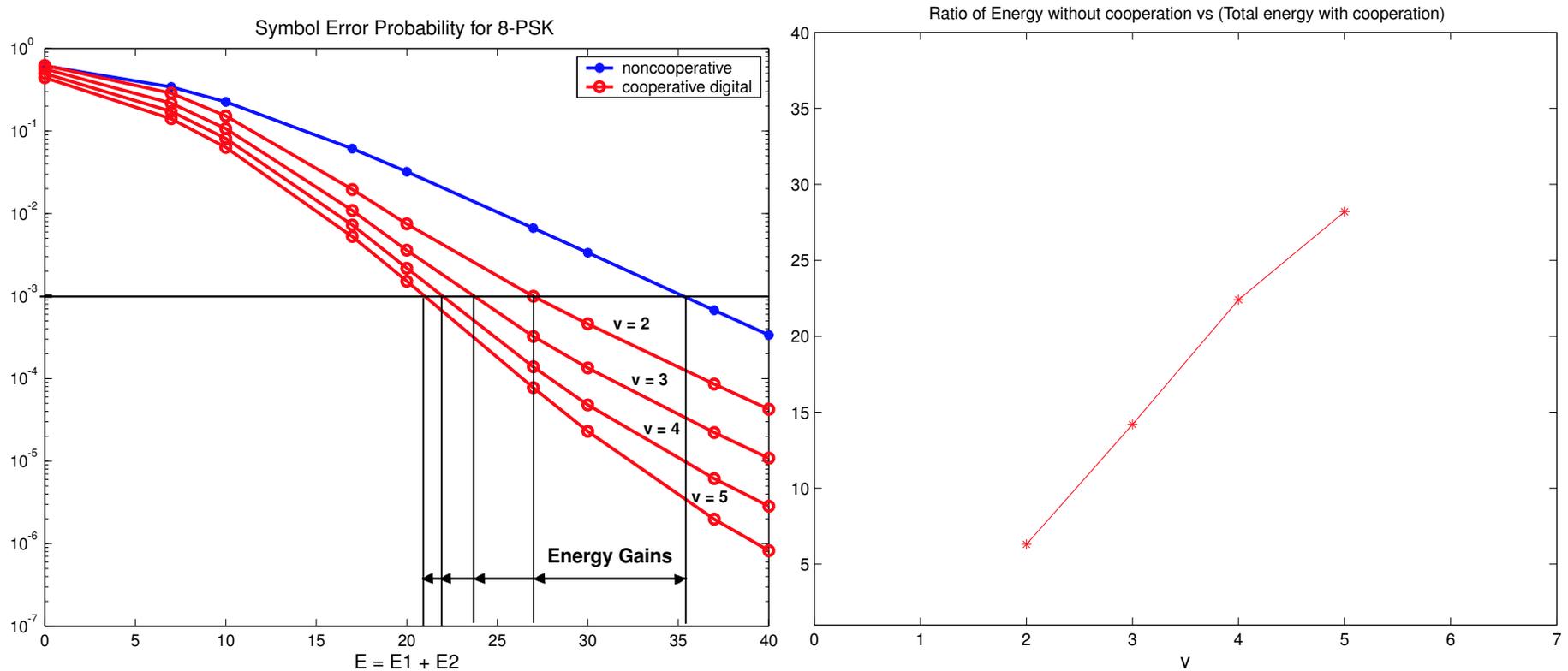
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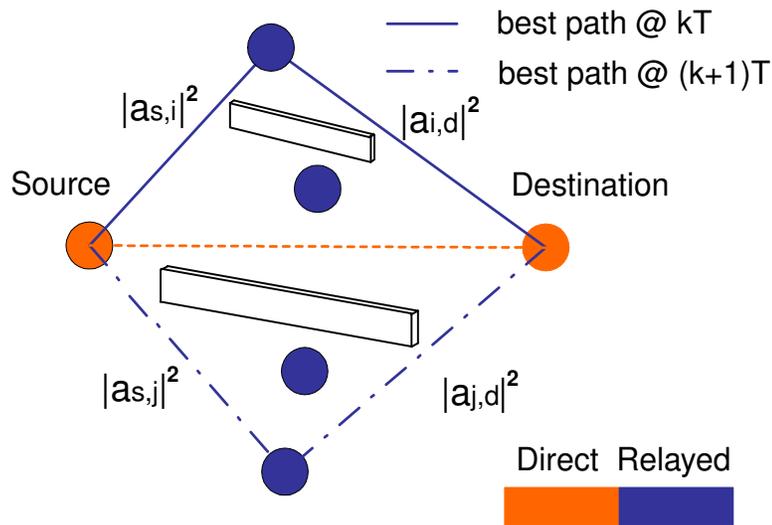
Results: Transmission Energy Gains



- Energy gains counterbalance the decrease of rate by a factor of 2.
- For the example above, 50% throughput increase is possible (8-PSK uncoded cooperative vs 2-PSK uncoded direct).

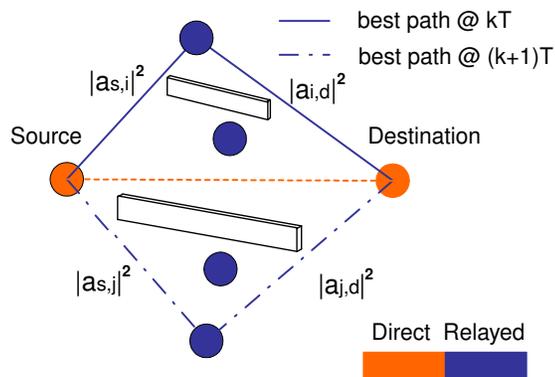
Results: Reception Energy Gains

- Cooperative reception of M relays \Rightarrow reception energy cost increases by a factor of M .
- Rx energy is comparable to Tx energy in modern radios [R. Min 2003].
- Proactive nature of Opportunistic Relaying, reception energy cost is fixed.



Results: Power Allocation Optimality (1)

- What if TOTAL power allocated to the relays was fixed?
- For amplify and forward networks, the equivalent system equation can be shown to be:
- It can be shown that opportunistic relaying is superior to other approaches in the field.

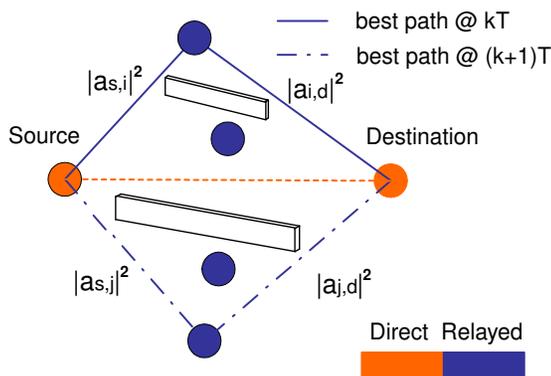


Results: Power Allocation Optimality (1)

- What if TOTAL power allocated to the relays was fixed?
- For amplify and forward networks, the equivalent system equation can be shown to be:

$$\begin{bmatrix} y_{D,1} \\ \frac{y_{D,2}}{\omega} \end{bmatrix} = \begin{bmatrix} \sqrt{P_{SD}} a_{SD} & 0 \\ \frac{1}{\omega} \sum_{i=1}^M \frac{\sqrt{P_{SRi}} \sqrt{P_{RiD}}}{\sqrt{P_{SRi} + N_0}} a_{SRi} a_{RiD} & \frac{1}{\omega} \sqrt{P_{SD}} a_{SD} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} n_{D,1} \\ \frac{\tilde{n}_{D,2}}{\omega} \end{bmatrix}$$

$$\mathcal{E}[\tilde{n}_{D,2} \tilde{n}_{D,2}^* | H_{R \rightarrow D}] = N_0 \underbrace{\left(1 + \sum_{i=1}^M \frac{P_{RiD} |a_{RiD}|^2}{P_{SRi} + N_0} \right)}_{\omega^2} = \omega^2 N_0 \quad (20)$$



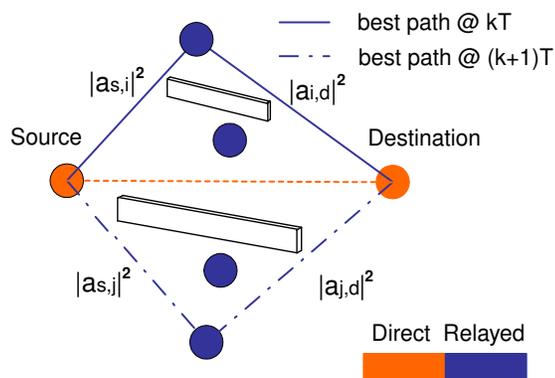
Results: Power Allocation Optimality (1)

- What if TOTAL power allocated to the relays was fixed?
- For amplify and forward networks, the equivalent system equation can be shown to be:

$$\mathbf{y} = \begin{bmatrix} \sqrt{P_{SD}} h_{SD} & 0 \\ H_{21} & \frac{1}{\omega} \sqrt{P_{SD}} h_{SD} \end{bmatrix} \mathbf{x} + \mathbf{n} \quad (21)$$

$$\mathbf{y} = \mathbf{H} \mathbf{x} + \mathbf{n} \quad (22)$$

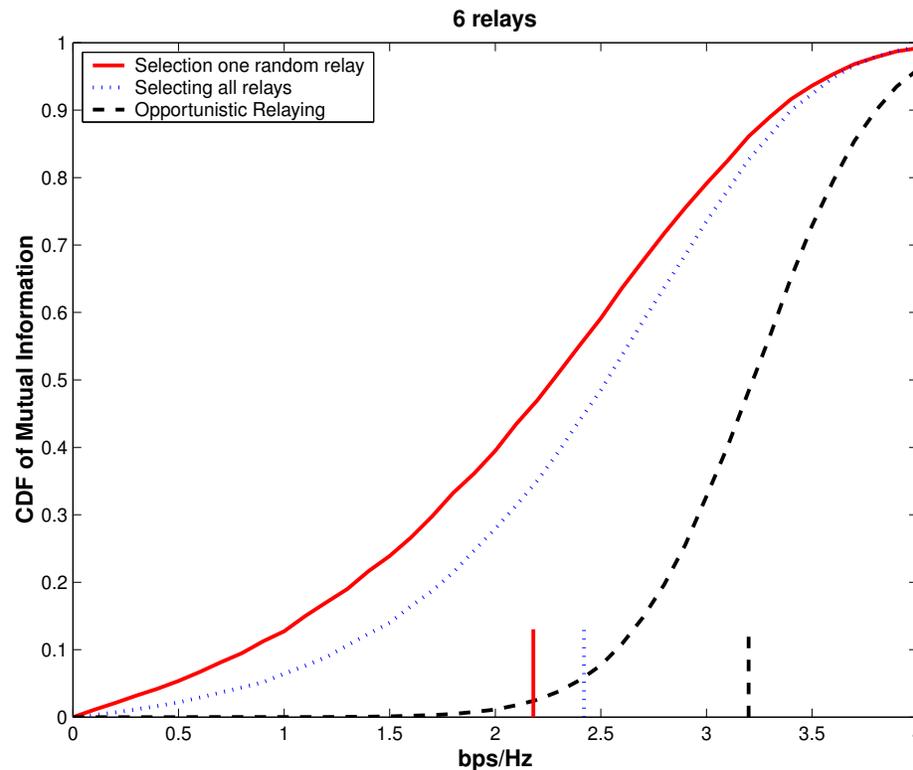
$$I_{AF} = \frac{1}{2} \log_2 \left(1 + \frac{P_{SD}}{N_0} |h_{SD}|^2 + \frac{|H_{21}|^2}{N_0} \right) \quad (23)$$



Results: Power Allocation Optimality (2)

Three cases considered, with all relays equivalent (same AVERAGE received SNR) :

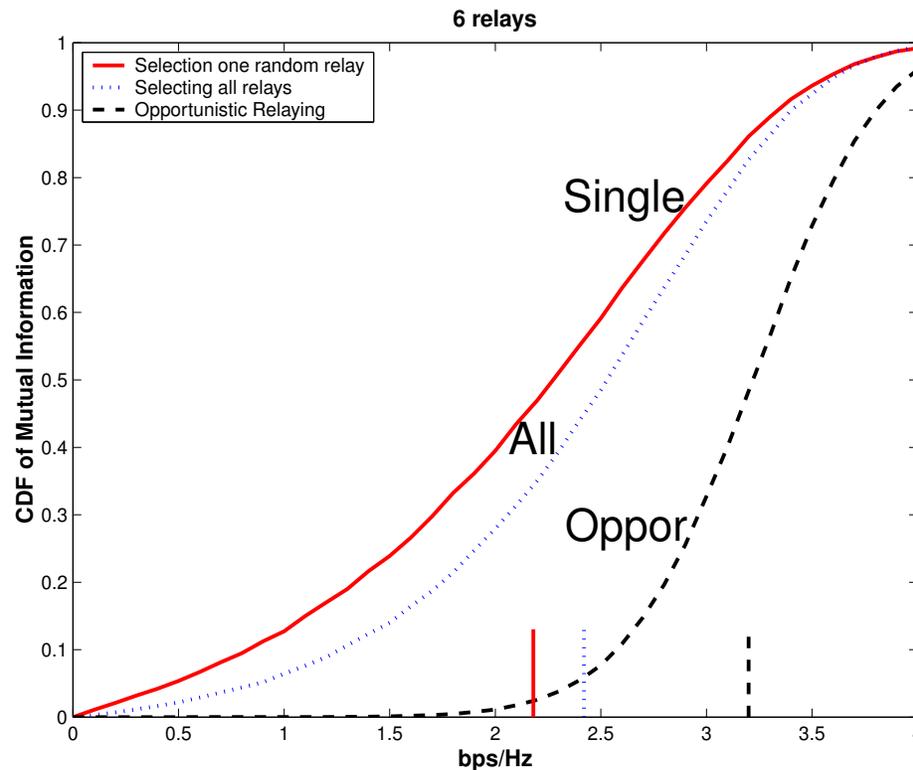
- Power to one relay (selection based on Average SNR).
- Power distributed to all relays (space-time coding).
- Power to opportunistic relay (Our Approach).



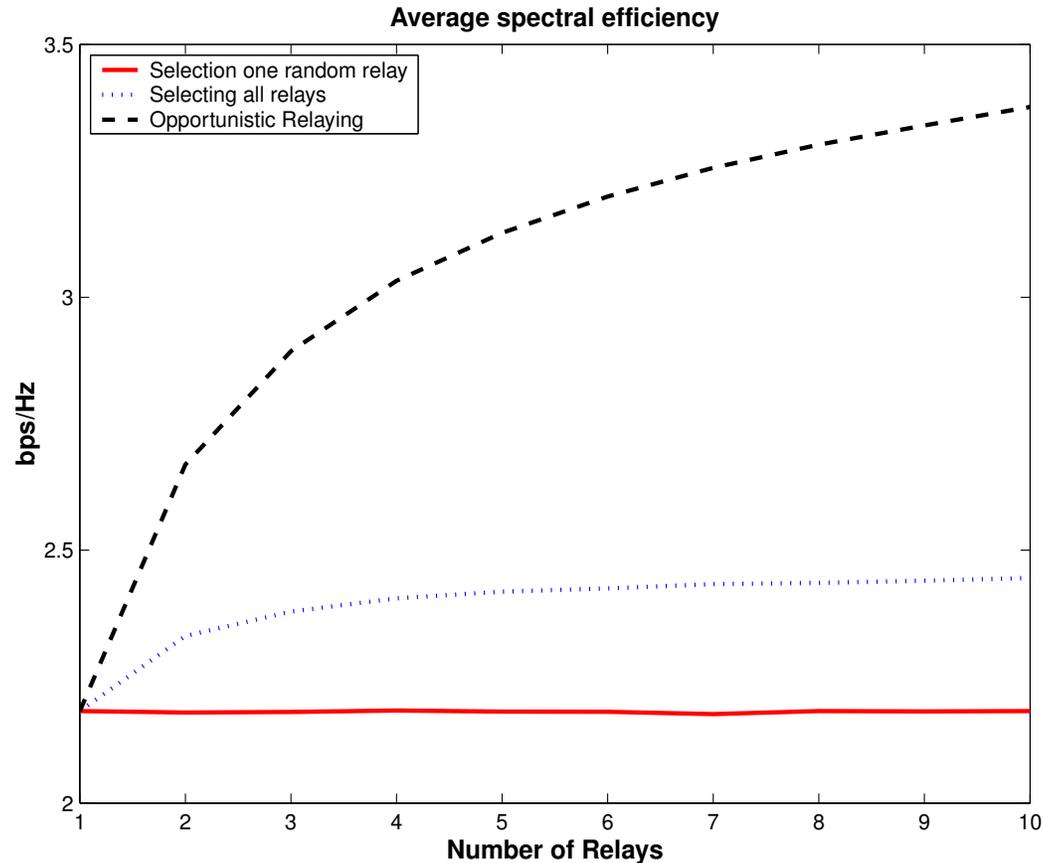
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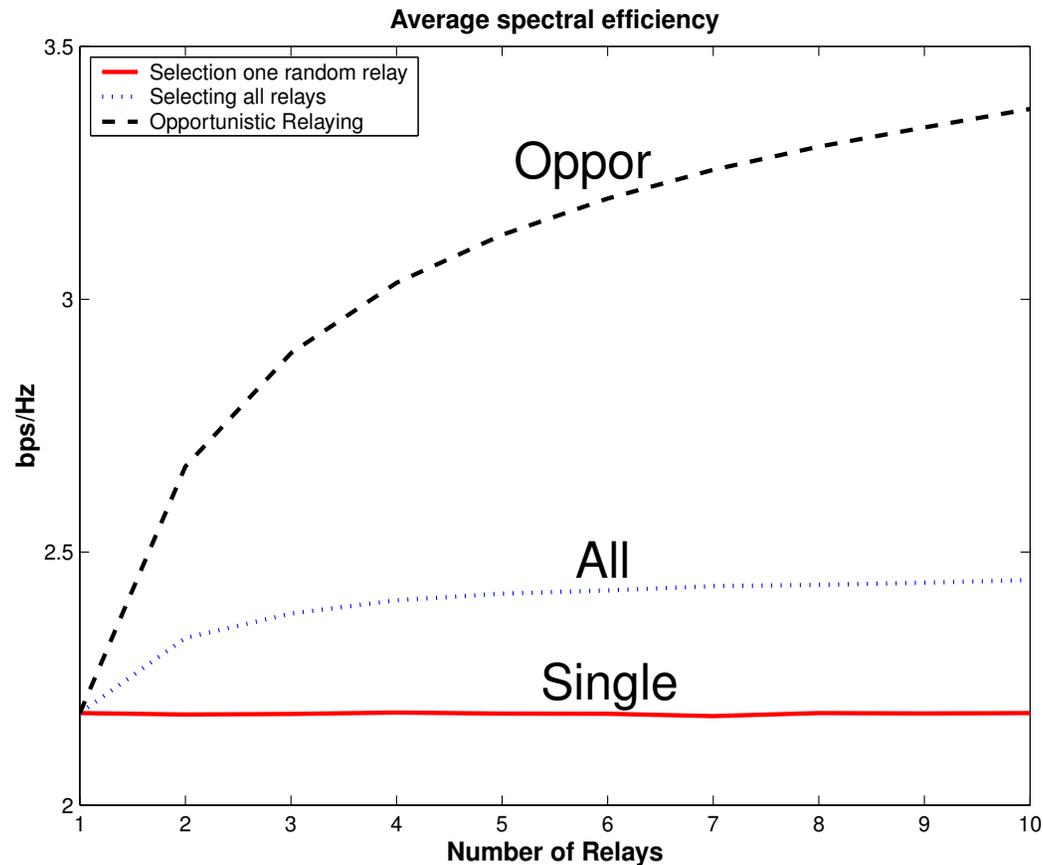


Results: Power Allocation Optimality (3)



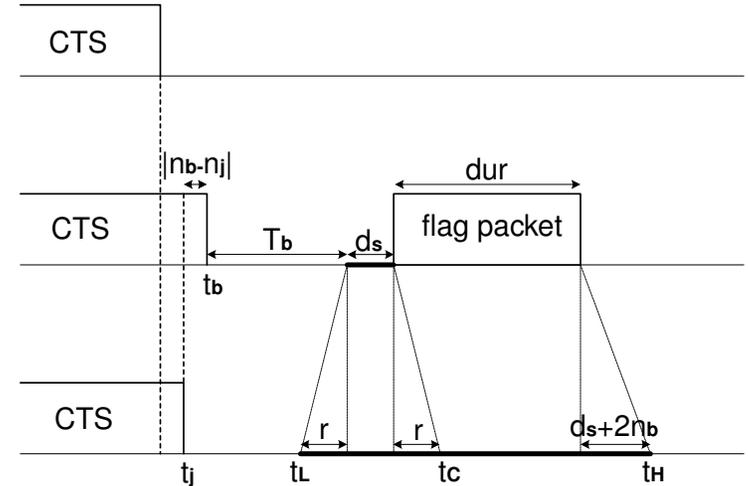
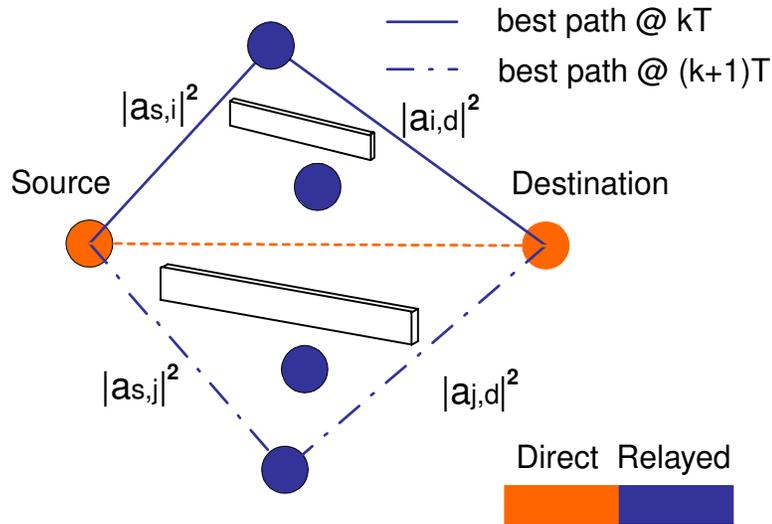
- Under a sum power constraint (and no beamforming capabilities) using all relays is suboptimal compared to opportunistic relaying.
- Similar results for decode and forward.

Results: Power Allocation Optimality (3)



- Under a sum power constraint (and no beamforming capabilities) using all relays is suboptimal compared to opportunistic relaying.
- Similar results for decode and forward.

Overhead: Collision Probability (1)



$$Policy I : h_i = \min\{|a_{si}|^2, |a_{id}|^2\} \quad Policy II : h_i = \frac{2}{\frac{1}{|a_{si}|^2} + \frac{1}{|a_{id}|^2}} = \frac{2 |a_{si}|^2 |a_{id}|^2}{|a_{si}|^2 + |a_{id}|^2}$$

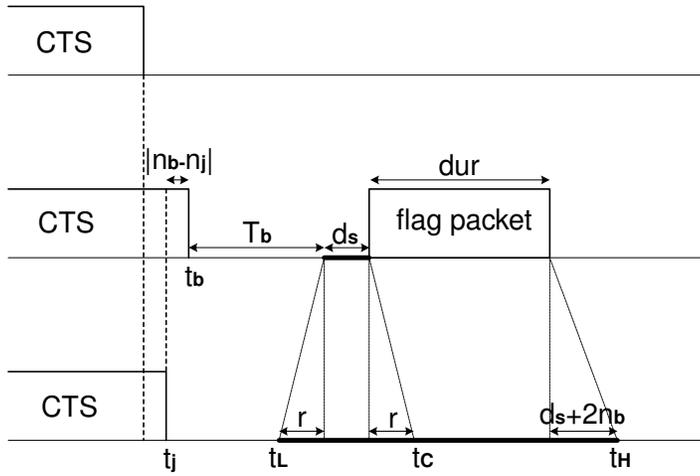
$$T_i = \frac{\lambda}{h_i} \quad (24)$$

Here λ has the units of time. For the discussion in this work, λ has simply values of $\mu secs$.

$$h_b = \max\{h_i\}, \iff \quad (25)$$

$$T_b = \min\{T_i\}, i \in [1..M]. \quad (26)$$

Overhead: Collision Probability (2)



Worst case scenario:

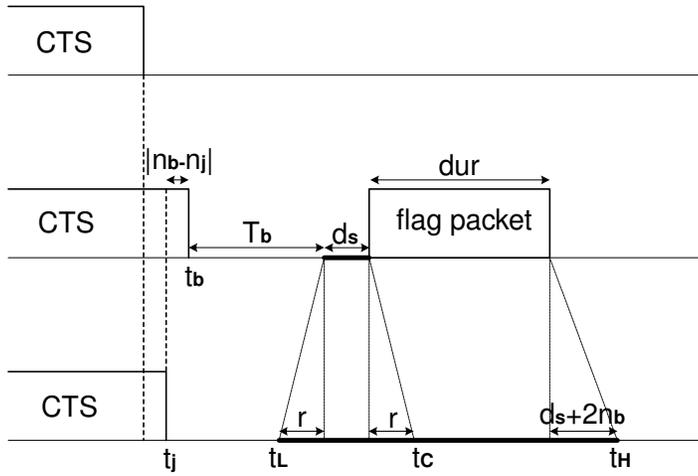
$$Pr(\text{Collision}) \leq Pr(\text{any } T_j < T_b + c \mid j \neq b)$$

where $T_b = \min\{T_j\}, j \in [1, M]$ and $c > 0$.

(a) *No Hidden Relays* : $c = r_{max} + |n_b - n_j|_{max} + d_s$

(b) *Hidden Relays* : $c = r_{max} + |n_b - n_j|_{max} + 2d_s + dur + 2n_{max}$

Overhead: Collision Probability (2)



Worst case scenario:

$$Pr(\text{Collision}) \leq Pr(\text{any } T_j < T_b + c \mid j \neq b)$$

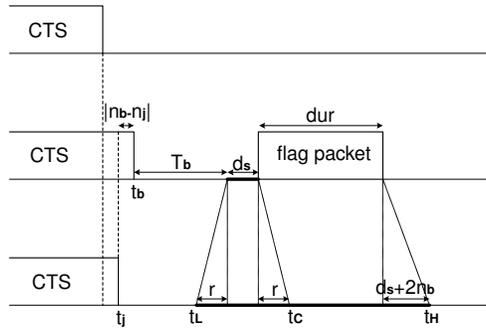
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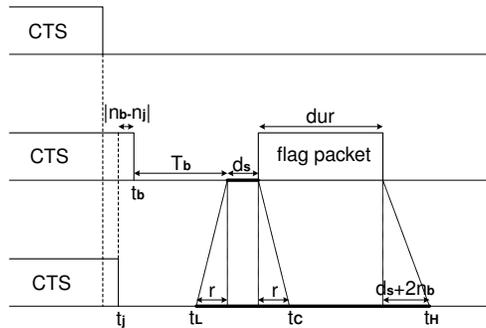
- n_j : propagation delay between relay j and destination. n_{max} is the maximum.
- r : propagation delay between two relays. r_{max} is the maximum.
- d_s : receive-to-transmit switch time of each radio.
- dur : duration of flag packet, transmitted by the "best" relay.

Overhead: Collision Probability (3)



If $T_b = \min\{T_j\}, j \in [1, M]$ and $Y_1 < Y_2 < \dots < Y_M$ the ordered random variables $\{T_j\}$ with $T_b \equiv Y_1$, and Y_2 the second minimum timer, then:

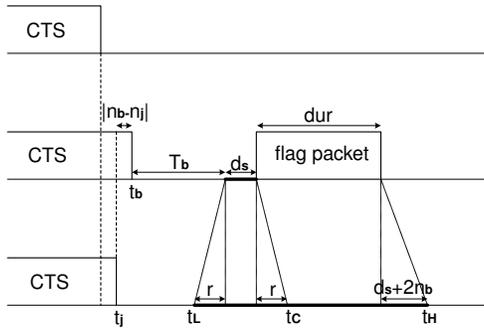
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$$Pr(\text{any } T_j < T_b + c \mid j \neq b) \equiv Pr(Y_2 < Y_1 + c) \quad (30)$$

Overhead: Collision Probability (3)



If $T_b = \min\{T_j\}, j \in [1, M]$ and $Y_1 < Y_2 < \dots < Y_M$ the ordered random variables $\{T_j\}$ with $T_b \equiv Y_1$, and Y_2 the second minimum timer, then:

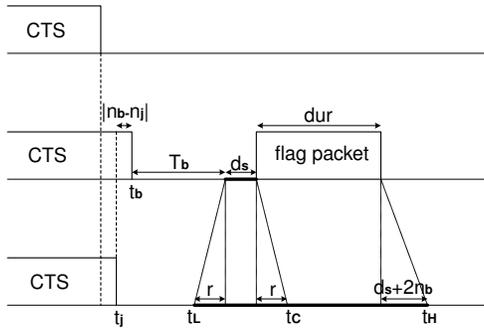
$$Pr(\text{any } T_j < T_b + c \mid j \neq b) \equiv Pr(Y_2 < Y_1 + c) \quad (33)$$

Given that $Y_j = \lambda/h_{(j)}$, $Y_1 < Y_2 < \dots < Y_M$ is equivalent to $1/h_{(1)} < 1/h_{(2)} < \dots < 1/h_{(M)}$

$$Pr(Y_2 < Y_1 + c) = Pr\left(\frac{1}{h_{(2)}} < \frac{1}{h_{(1)}} + \frac{c}{\lambda}\right) \quad (34)$$

Ratio $\frac{\lambda}{c}$ needs to be as high as possible. λ and c are user controlled.

Overhead: Collision Probability (3)



If $T_b = \min\{T_j\}, j \in [1, M]$ and $Y_1 < Y_2 < \dots < Y_M$ the ordered random variables $\{T_j\}$ with $T_b \equiv Y_1$, and Y_2 the second minimum timer, then:

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$$Pr(Y_2 < Y_1 + c) = Pr\left(\frac{1}{h_{(2)}} < \frac{1}{h_{(1)}} + \frac{c}{\lambda}\right) \quad (37)$$

Ratio $\frac{\lambda}{c}$ needs to be as high as possible. λ and c are user controlled.

However λ needs to be kept small:

$$E[T_j] = E[\lambda/h_j] \geq \lambda/E[h_j] \quad (38)$$

Overhead: Collision Probability (4)

Lemma: Given $M \geq 2$ i.i.d. positive random variables T_1, T_2, \dots, T_M , each with probability density function $f(x)$ and cumulative distribution function $F(x)$, and $Y_1 < Y_2 < Y_3 \dots < Y_M$ are the M ordered random variables T_1, T_2, \dots, T_M , then $\Pr(Y_2 < Y_1 + c)$, where $c > 0$, is given by the following equations:

$$\Pr(Y_2 < Y_1 + c) = 1 - I_c \quad (39)$$

$$I_c = M(M-1) \int_c^{+\infty} f(y) [1 - F(y)]^{M-2} F(y-c) dy \quad (40)$$

• Wireless channel statistics of $h \Rightarrow$ pdf f and cdf F of $T = \lambda/h \Rightarrow \Pr(\text{collision})$.

Overhead: Collision Probability (4)

Lemma: Given $M \geq 2$ i.i.d. positive random variables T_1, T_2, \dots, T_M , each with probability density function $f(x)$ and cumulative distribution function $F(x)$, and $Y_1 < Y_2 < Y_3 \dots < Y_M$ are the M ordered random variables T_1, T_2, \dots, T_M , then $Pr(Y_2 < Y_1 + c)$, where $c > 0$, is given by the following equations:

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$$I_c = M (M - 1) \int_c^{+\infty} f(y) [1 - F(y)]^{M-2} F(y - c) dy \quad (42)$$

- Wireless channel statistics of $h \Rightarrow$ pdf f and cdf F of $T = \lambda/h \Rightarrow Pr(\text{collision})$.
- Example: for a mobility of $0 - 3$ km/h \Rightarrow maximum Doppler shift is $f_m = 2.5$ Hz \Rightarrow minimum coherence time on the order of $T_c \simeq 200$ milliseconds.

For $c/\lambda \approx 1/200 \Rightarrow Pr(\text{Collision}) \leq 0.6\%$ for policy I.

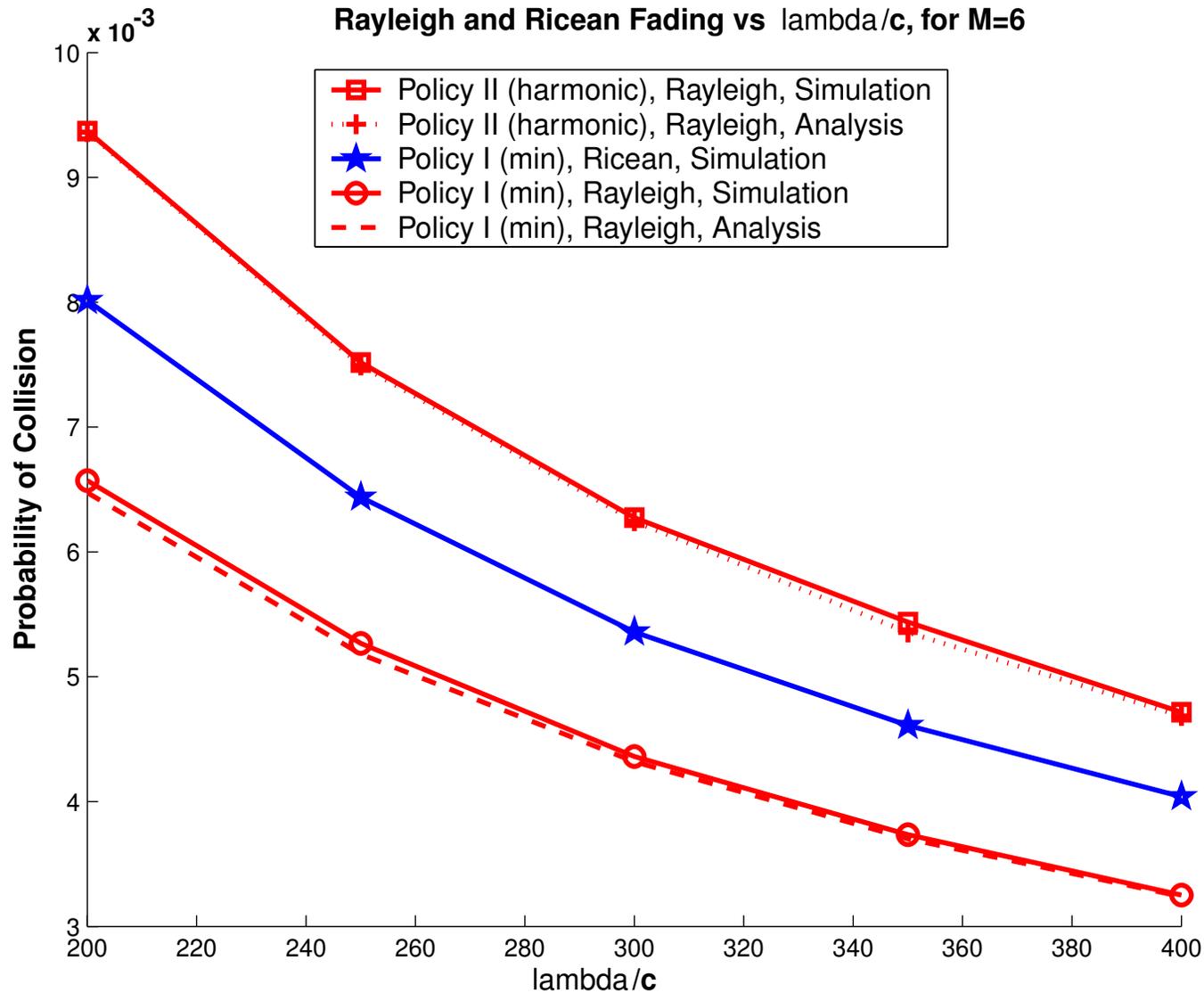
For $c \approx 5\mu s \Rightarrow \lambda \approx 1ms \simeq \frac{1}{100} T_c$.

For $c \approx 1\mu s \Rightarrow \lambda \approx 200\mu s \simeq \frac{1}{1000} T_c$.

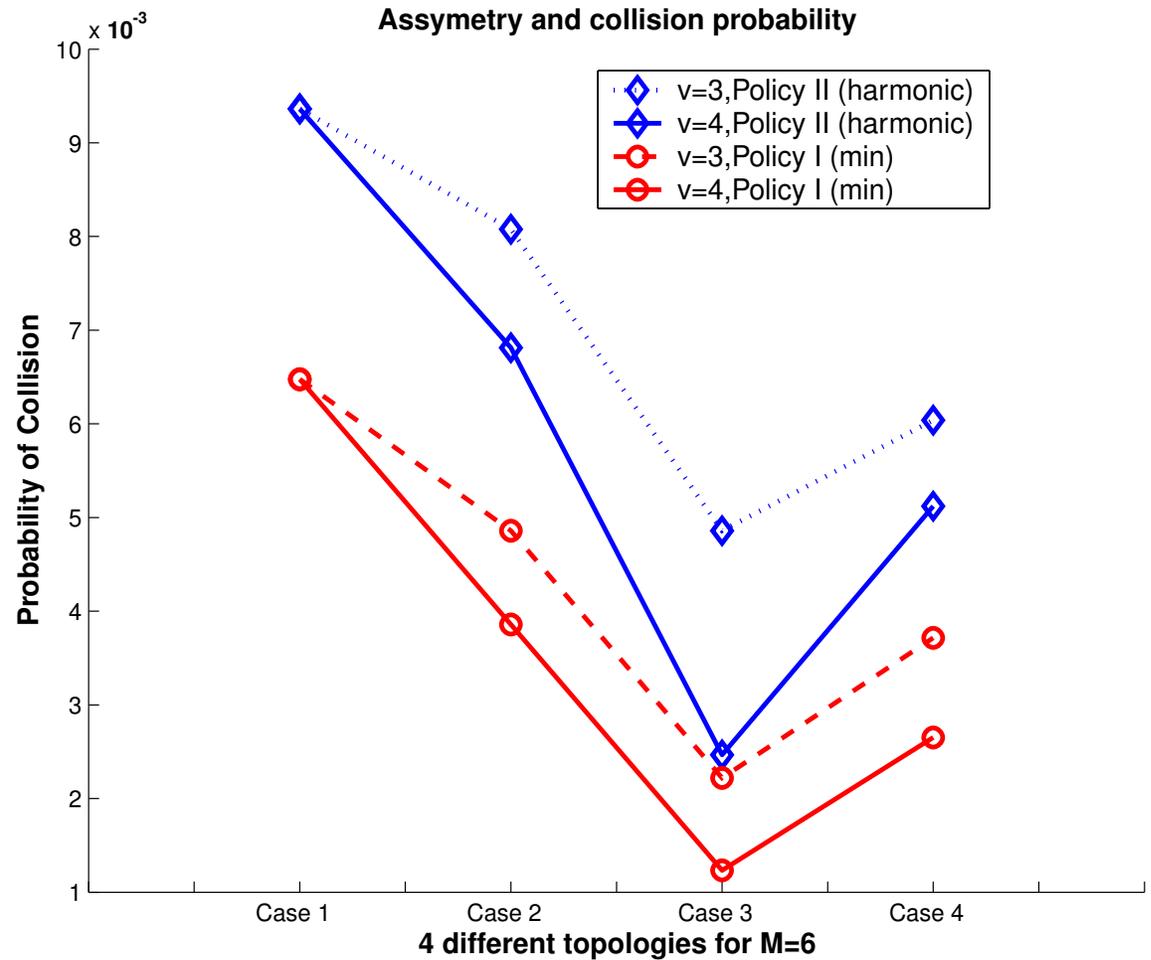
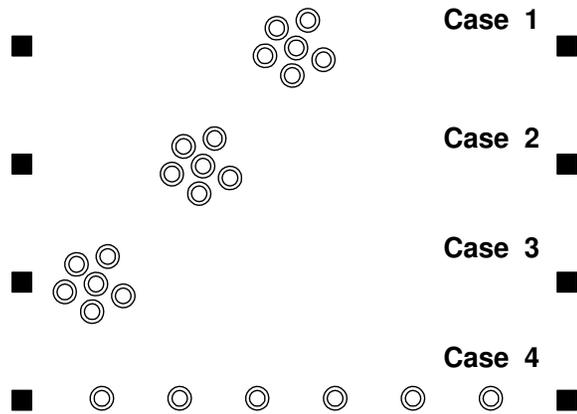
Rigorous analysis earns you trips around the world...



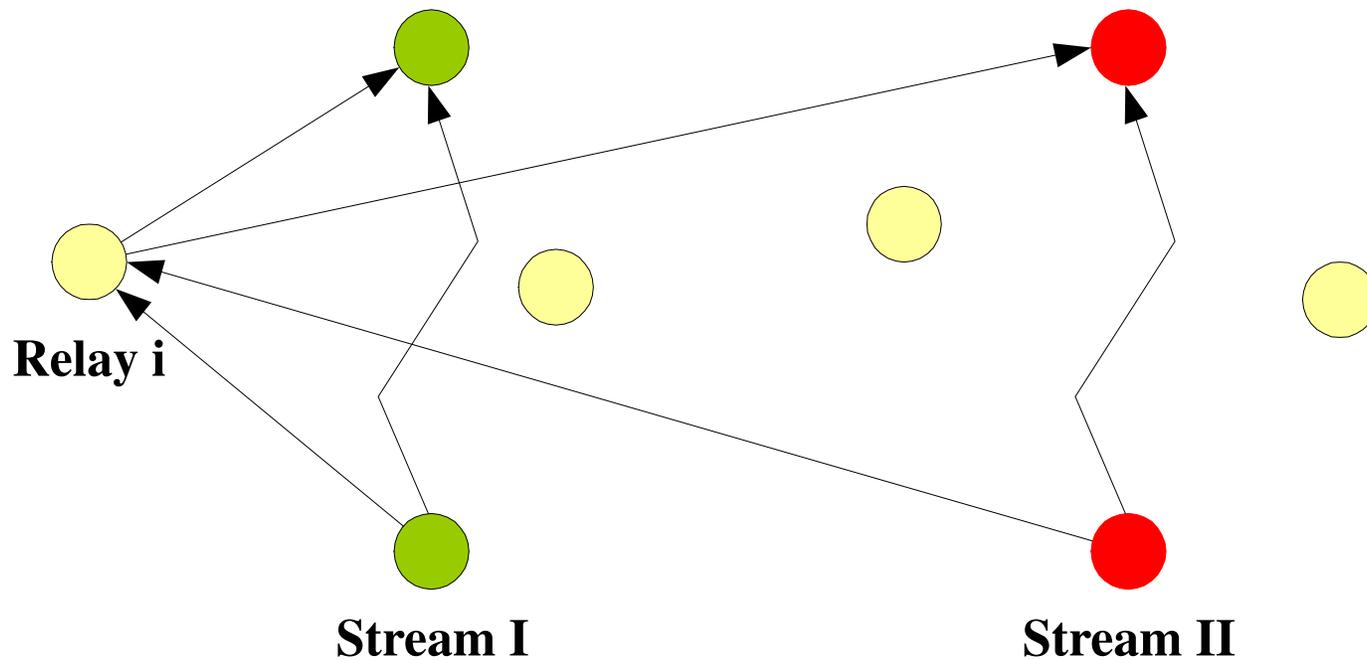
Overhead: Collision Probability (5)



Overhead: Collision Probability (6)

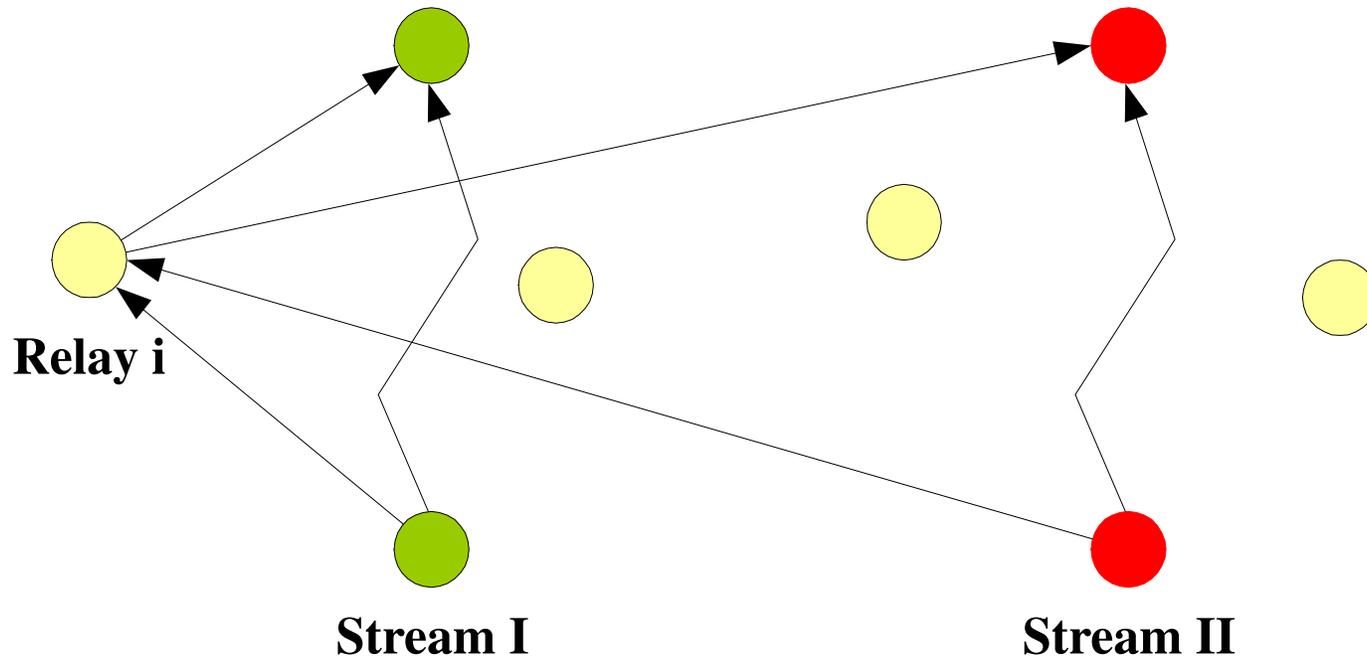


...and a Remark...



$$b = \underbrace{\arg \max}_i \{ \min \{ SNR_{si}, SNR_{id} \} \} = \max \{ SNR_{sid} \}, \quad i \in [1..M] \quad (43)$$

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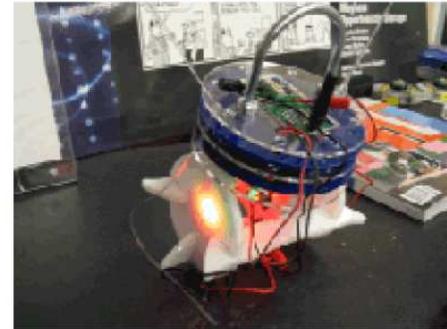
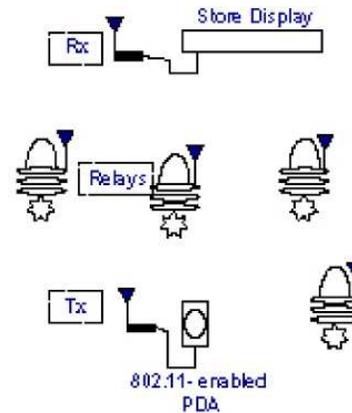
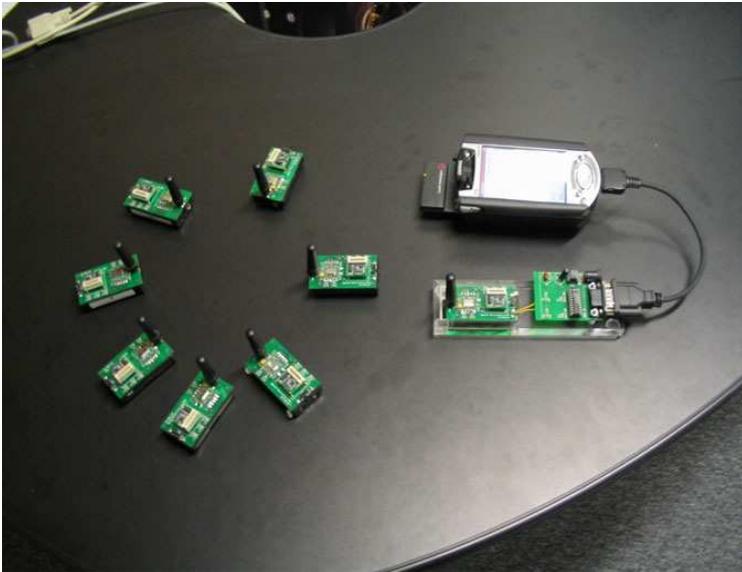
$$b = \underbrace{\arg \max}_i \{ \min \{ SNR_{si}, SNR_{id} \} \} = \max \{ SNR_{sid} \}, \quad i \in [1..M] \quad (45)$$

$$b = \underbrace{\arg \max}_i \{ \min \{ SINR_{si}, SINR_{id} \} \} = \max \{ SINR_{sid} \}, \quad i \in [1..M] \quad (46)$$

Outline

- Assumptions and Background
- Approach
- Performance
- **Implementation Example**
- Relevant Technologies
- Conclusion
- Acknowledgements

Implementation: Hardware



- Rethinking wireless:
approach needs access to physical (layer 1), link (layer 2), routing (layer 3).
- COTS radios usually give limited access to all layers ⇒
- We built our own low cost embedded Software Defined Radios (SDRs).
- We built a room size cooperative diversity demo.

Implementation: Demo Setup

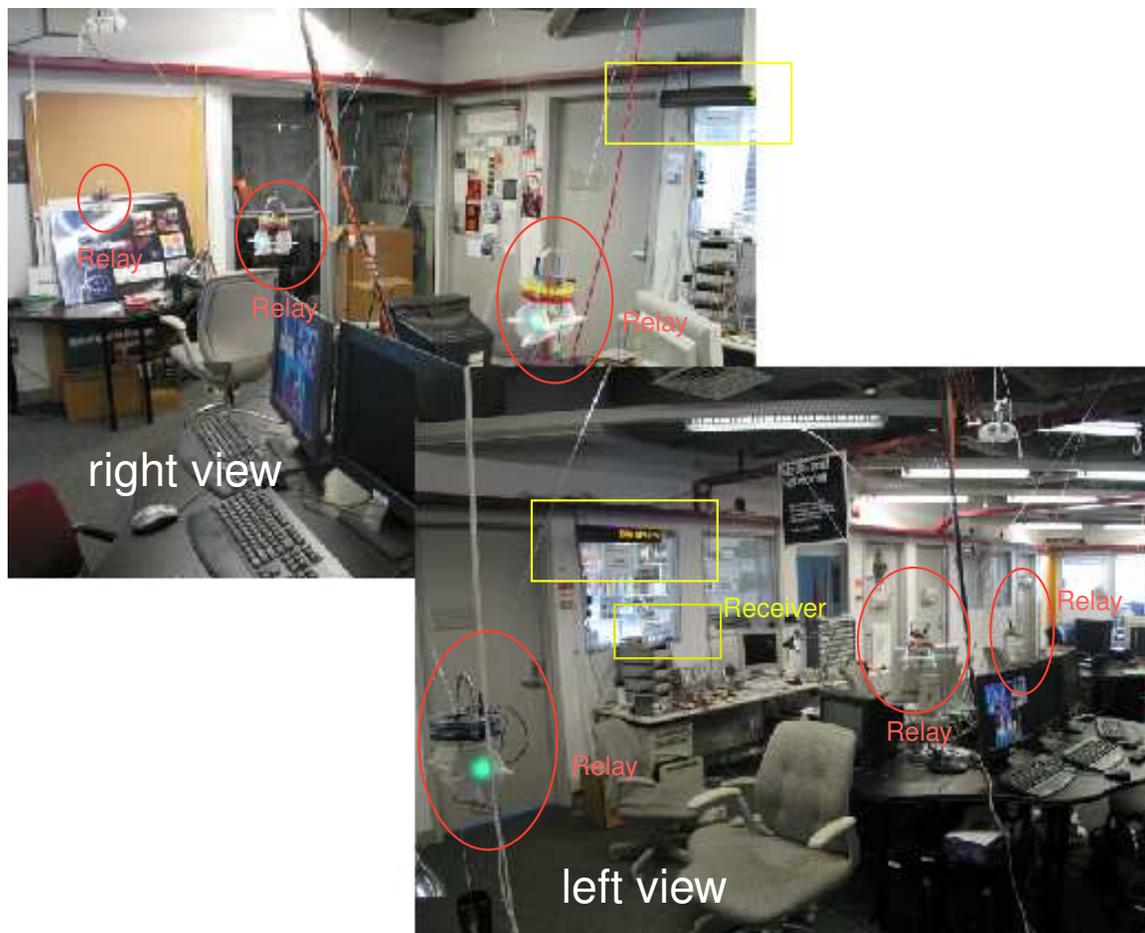


right view

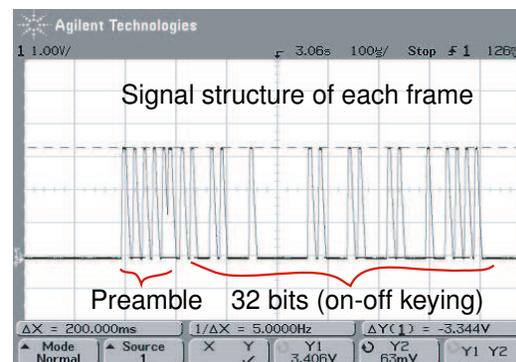
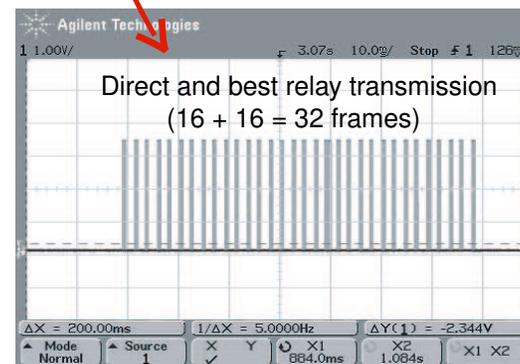
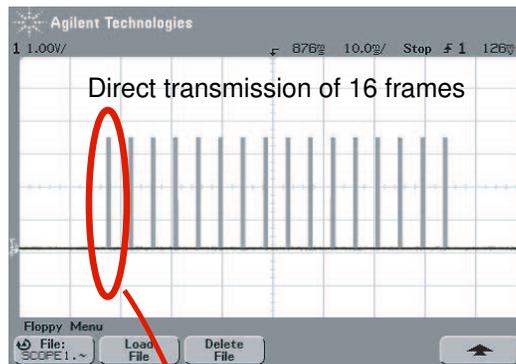
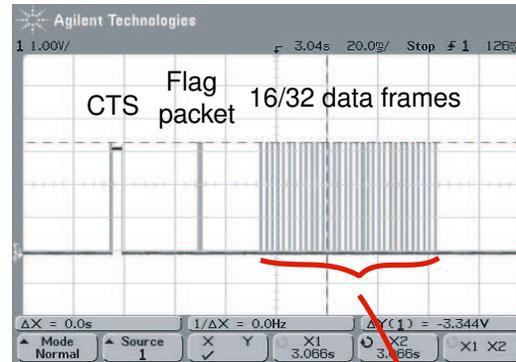


left view

Implementation: Demo Setup



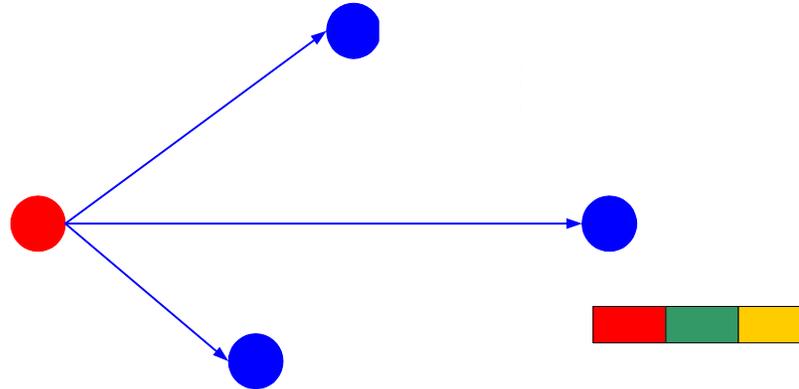
Implementation: Signal Structure



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Coordination, Cooperation and Time Keeping

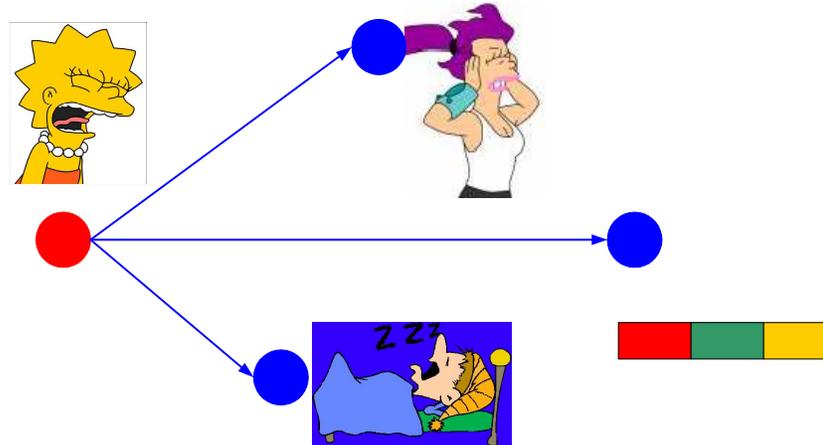


- Relays (or receiver) might be busy or in sleep mode!
- Time keeping could simplify required *scheduling*.
- *Time keeping as the basis of scalable communication.*

Extensive work on Network Time Keeping:

- centralized
- decentralized

Coordination, Cooperation and Time Keeping

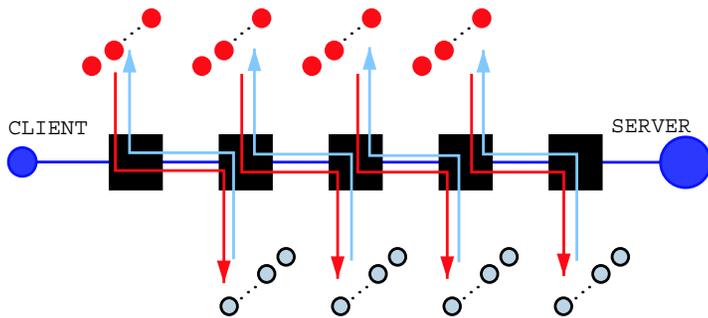


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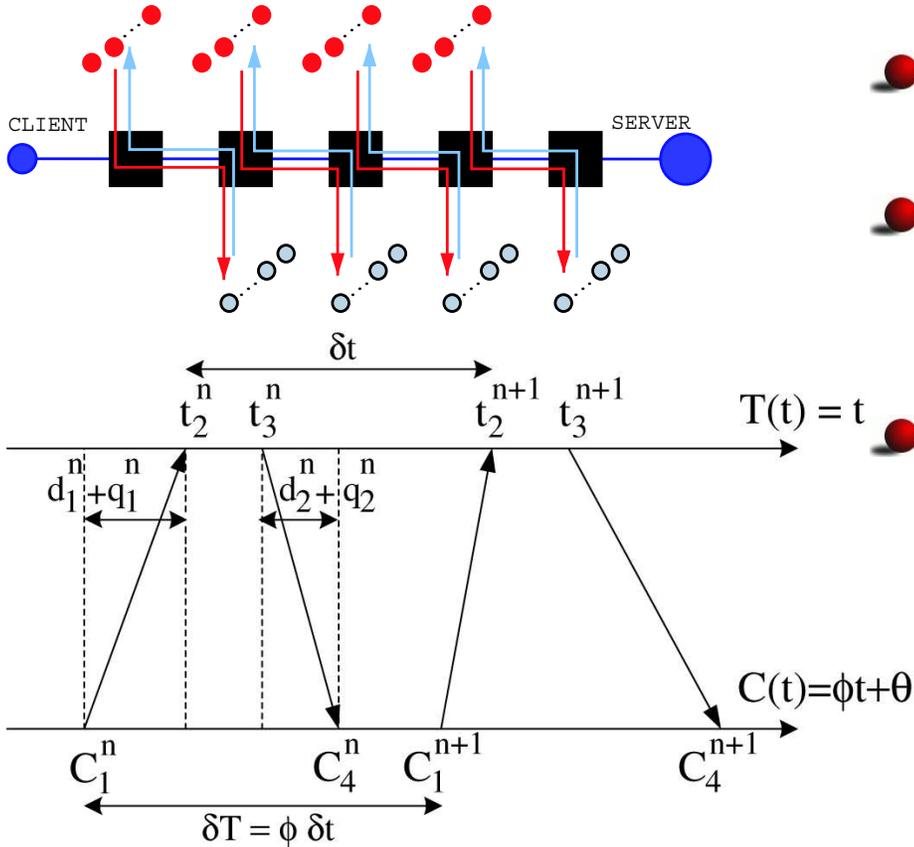
Centralized Time Keeping



- No control over the network: noisy environment.
- No control over the time server: would like to use existing infrastructure.
- Three End-to-End algorithms were compared:
 - Averaging (NIST).
 - Linear Programming (proposed before).
 - Kalman Filtering (our proposal).

Estimation of ϕ and θ , with minimum communication BW and computation requirements.

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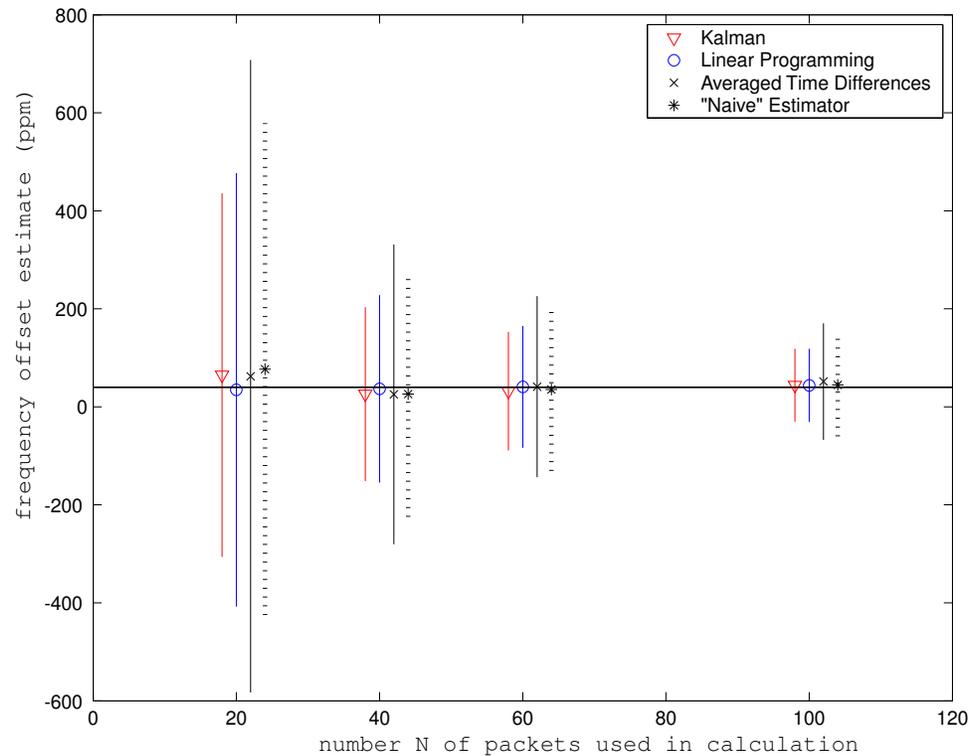
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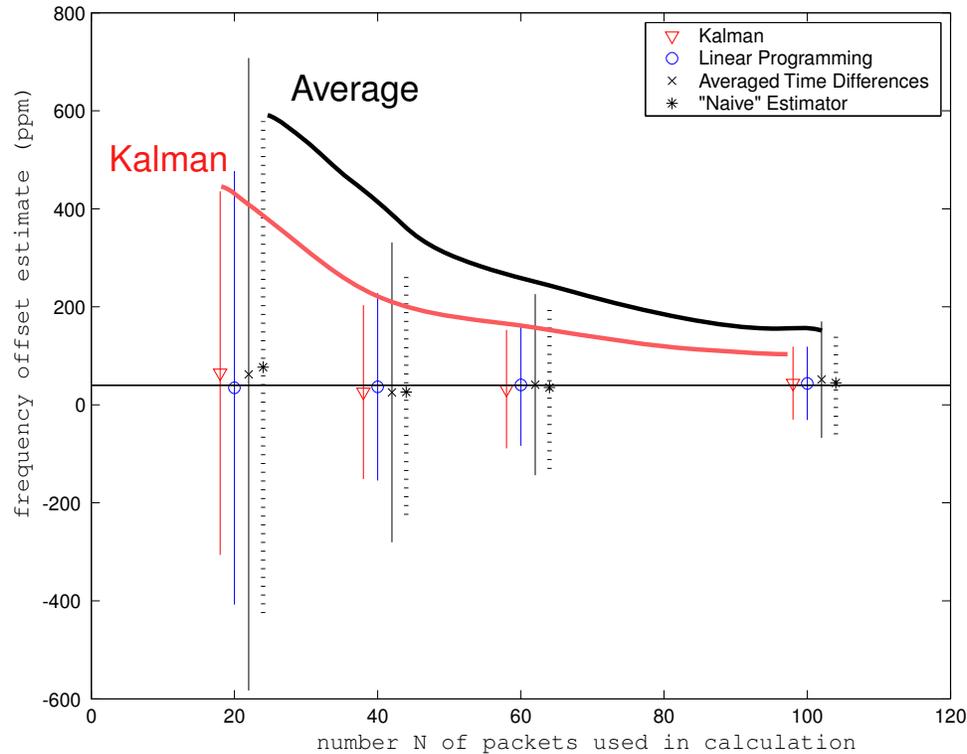
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Centralized Time Keeping Results



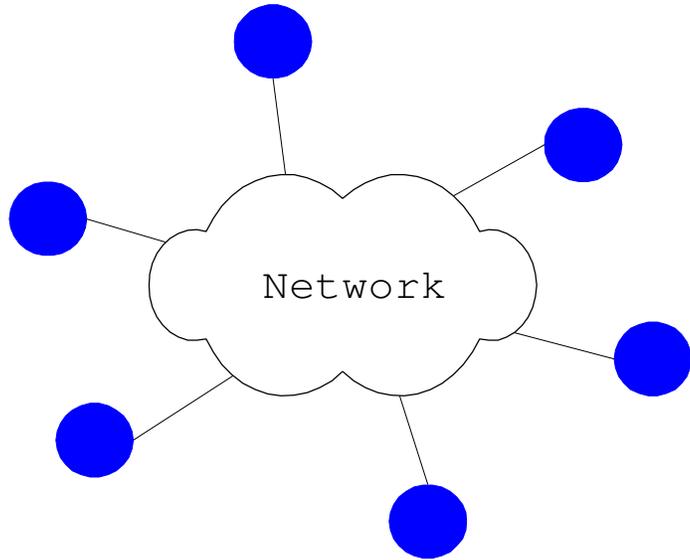
- Improving *accuracy* (error) and *precision* (variance of error), compared to existing approaches.
- Computation efficient (since it is recursive) -
- Implemented and tested using existed NTP infrastructure.

Centralized Time Keeping Results



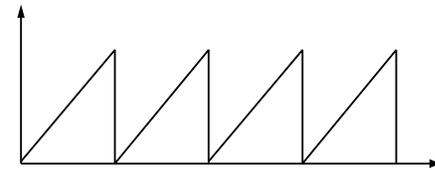
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Decentralized Time Keeping

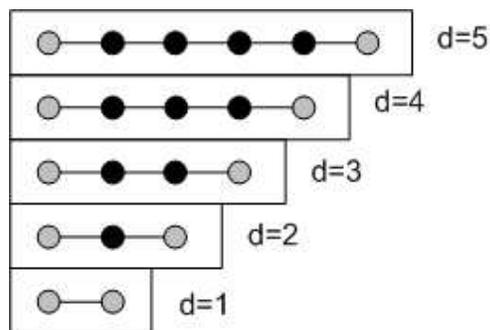


The network is the time server.

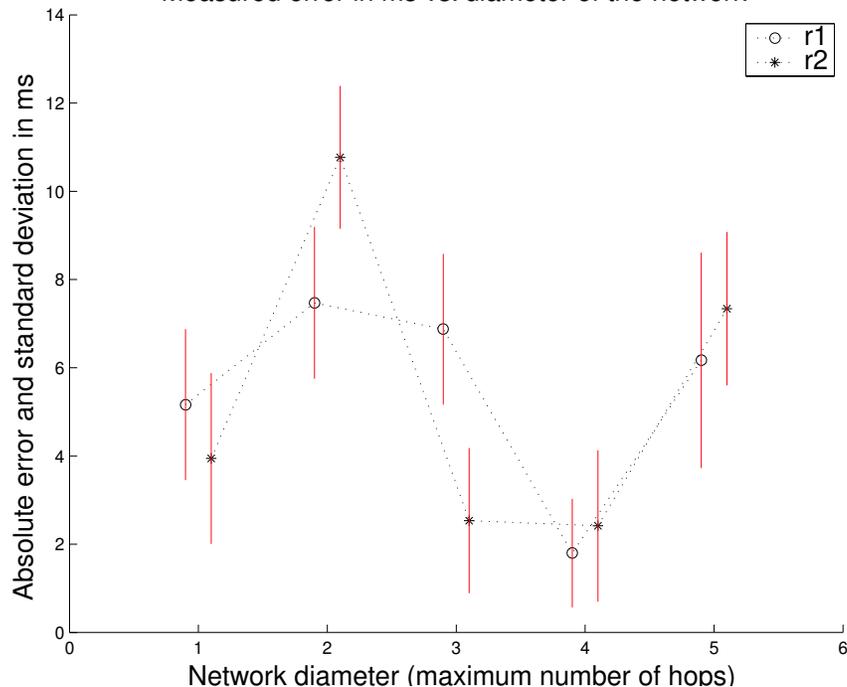
- Only local communication.
- Exchange timestamps and keep the highest (Lamport's idea).
- Redefine time as a periodic function!
- The network *re-calibrates* periodically and autonomously.



Decentralized Time Keeping Results



Measured error in ms vs. diameter of the network



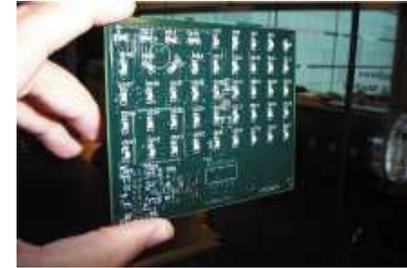
- Error could decrease with increasing Network diameter!

$$\begin{aligned} \epsilon(t_c) &= C_i(t_c) - C_j(t_c) = \\ &= \epsilon(t_0 + x) + (\phi_i - \phi_j) \Delta t \\ \Delta t &= t_c - (t_0 + x) \end{aligned}$$

- Error depends on communication BW.

$$x = \text{propagation delay} + \text{transmission delay} + \text{operating system delay.}$$

Decentralized Time Keeping Demo



- Objective: play music in synchrony, display *heartbeat* at the edges...

Decentralized Time Keeping Demo (2)



- This algorithm is based on oscillator's coupling (no averaging).
- Coupling among terminals with semi-periodic signal \equiv *Entrainment*.
- It is relevant to natural phenomena of synchronization (fireflies, cardiac neurons etc.)

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- Implementation Example
- Relevant Technologies
- Conclusion
- Acknowledgements

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Papers

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